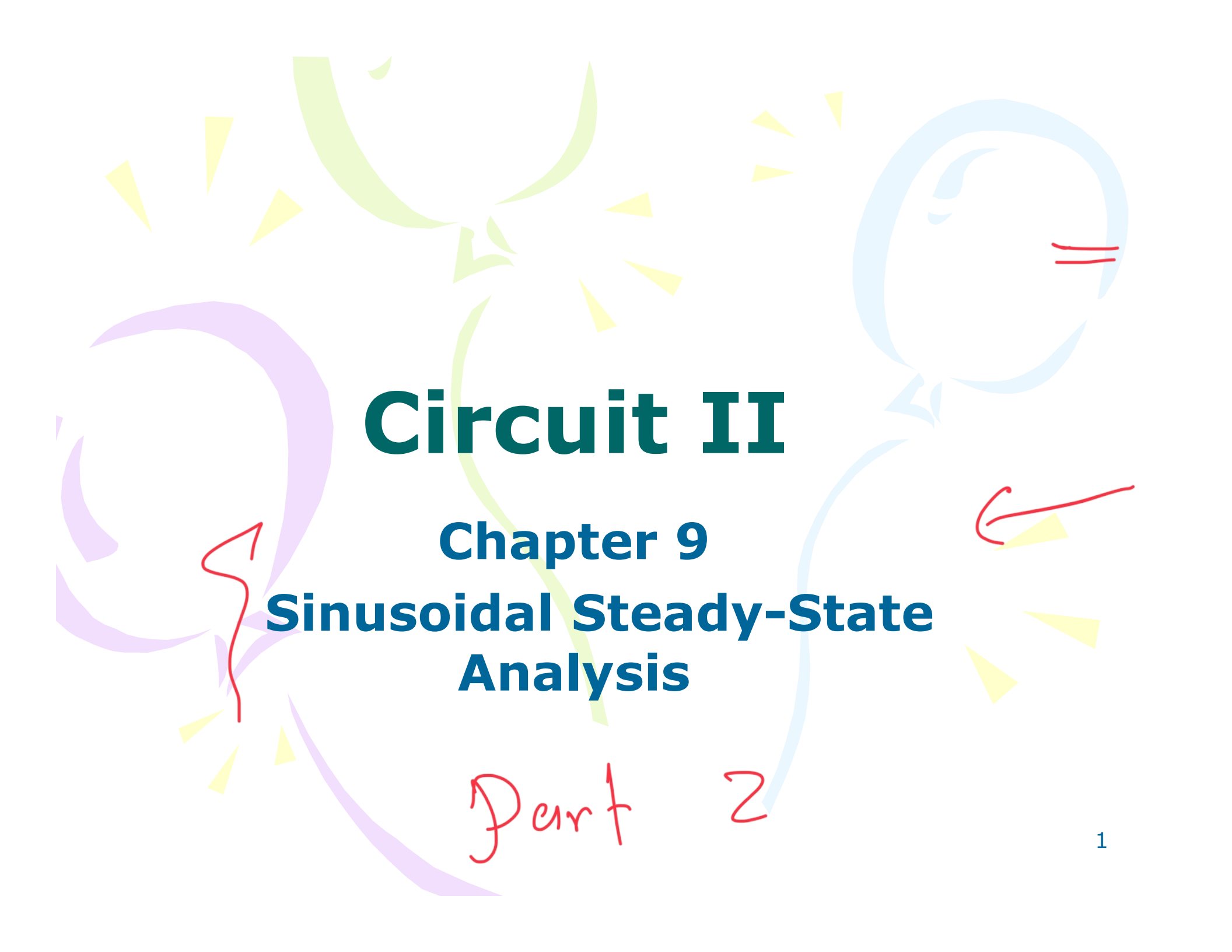




Circuit II

Chapter 9 Sinusoidal Steady-State Analysis

Part 2

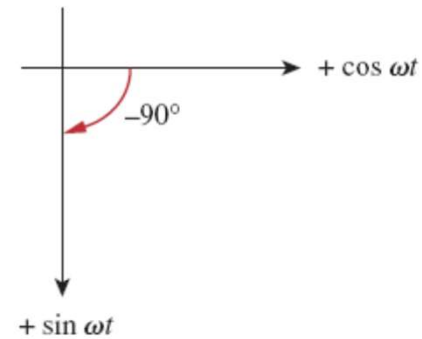
We can transform a sinusoid from sine form to cosine form or vice versa.

$$\sin(\omega t \pm 180^\circ) = -\sin \omega t$$

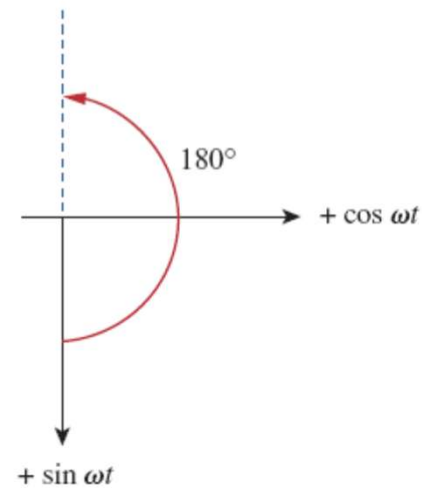
$$\cos(\omega t \pm 180^\circ) = -\cos \omega t$$

$$\sin(\omega t \pm 90^\circ) = \pm \cos \omega t$$

$$\cos(\omega t \pm 90^\circ) = \mp \sin \omega t$$



(a)



(b)

Figure 9.3

A graphical means of relating cosine and sine: (a) $\cos(\omega t - 90^\circ) = \sin \omega t$, (b) $\sin(\omega t + 180^\circ) = -\sin \omega t$.

$$\begin{aligned} \ominus \sin(\omega t) &\rightarrow \boxed{+} \sin(\omega t \pm 180^\circ) \quad \boxed{1} \\ \ominus \cos(\omega t) &\rightarrow \boxed{+} \cos(\omega t \pm 180^\circ) \end{aligned}$$

± لا تفرق بين الزاوية الزاوية
 $\rightarrow \underline{180^\circ} \text{ mod } 360^\circ$

Ex $\ominus \sin(\omega t + 40^\circ)$

$$+ \sin(\omega t + 40^\circ \pm 180^\circ)$$

$$\boxed{+ \sin(\omega t - 140^\circ)}$$

$$40 + 180 \rightarrow 220^\circ$$

$$40 - 180 \rightarrow -140^\circ$$

220

$$\ominus \sin(\omega t) \longrightarrow \boxed{+} \cos(\omega t + 90^\circ)$$

$$\oplus \sin(\omega t) \longrightarrow \boxed{+} \cos(\omega t - 90^\circ)$$

just

$$\oplus \cos(\omega t) \longrightarrow \boxed{+} \sin(\omega t + 90^\circ)$$

$$\ominus \cos(\omega t) \longrightarrow \boxed{+} \sin(\omega t - 90^\circ)$$

to

so

9.2 Sinusoids (4)

Example 3

Find the phase angle between $i_1 = -4 \sin(377t + 25^\circ)$ and $i_2 = 5 \cos(377t - 40^\circ)$, does i_1 lead or lag i_2 ?

Solution:

Since $\sin(\omega t + 90^\circ) = \cos \omega t$

$$i_2 = 5 \sin(377t - 40^\circ + 90^\circ) = 5 \sin(377t + 50^\circ)$$

$$i_1 = -4 \sin(377t + 25^\circ) = 4 \sin(377t + 180^\circ + 25^\circ) = 4 \sin(377t + 205^\circ)$$

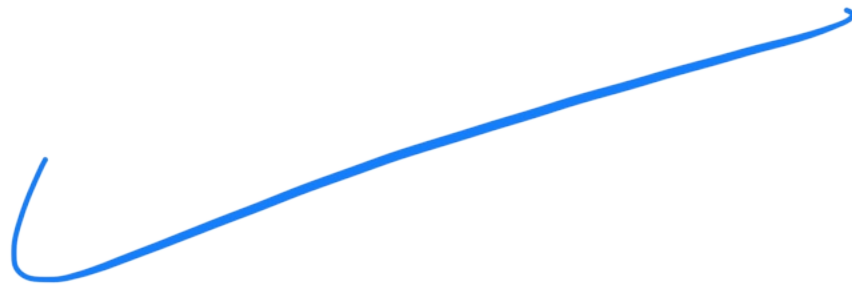
therefore, i_1 leads i_2 155° .

ملاحظة للتفريق، $\sin$ lead و $\cos$ lag

[1] أن يكون إزاحة موجية "منيرة"

[2] أن يكون إزاحة $\sin$ أو $\cos$

[3] أن يكون إزاحة $\sin$ أو $\cos$ مع التردد ω



$$i_1 = \textcircled{-} 4 \sin(377t + 25) \quad ??$$

$$i_2 = +5 \cos(377t - 40) \quad \checkmark$$

$$= 4 \sin(377t + 25)$$

$$\Rightarrow +4 \cos(377t + 25 + 90)$$

$$i_1 = 4 \cos(377t + 115) \quad \leftarrow \theta_1 \quad \checkmark \checkmark \checkmark$$

$$i_2 = 5 \cos(377t - 40) \quad \leftarrow \theta_2 \quad \checkmark \checkmark \checkmark$$

lead $\rightarrow$ القائد "متقدم"

lag $\rightarrow$ متأخر "الأخيرة"

$$\theta_1 = 115$$

$$\theta_1 > \theta_2$$

$$\theta_2 = -40$$

الأخير $\leftarrow$ القائد

C_1 lead C_2

"متقدم كل C_2 "

C_2 lag C_1

" C_2 متأخر على C_1 "

C_1 lead C_2

$$\theta_1 - \theta_2 = 115 - (-40)$$

$$115 + 40 = 155 \leftarrow$$

C_1 lead C_2 by 155

Example 4

Calculate the phase angle between $v_1 = -10 \cos(\omega t + 50^\circ)$ and $v_2 = 12 \sin(\omega t - 10^\circ)$. State which sinusoid is leading.

■ **METHOD 1** In order to compare v_1 and v_2 , we must express them in the same form. If we express them in cosine form with positive amplitudes,

$$v_1 = -10 \cos(\omega t + 50^\circ) = 10 \cos(\omega t + 50^\circ - 180^\circ)$$
$$v_1 = 10 \cos(\omega t - 130^\circ) \quad \text{or} \quad v_1 = 10 \cos(\omega t + 230^\circ) \quad (9.2.1)$$

and

$$v_2 = 12 \sin(\omega t - 10^\circ) = 12 \cos(\omega t - 10^\circ - 90^\circ)$$
$$v_2 = 12 \cos(\omega t - 100^\circ) \quad (9.2.2)$$

It can be deduced from Eqs. (9.2.1) and (9.2.2) that the phase difference between v_1 and v_2 is 30° . We can write v_2 as

$$v_2 = 12 \cos(\omega t - 130^\circ + 30^\circ) \quad \text{or} \quad v_2 = 12 \cos(\omega t + 260^\circ) \quad (9.2.3)$$

Comparing Eqs. (9.2.1) and (9.2.3) shows clearly that v_2 leads v_1 by 30° .

$$V_1 = \ominus 10 \cos(\omega t + 50)$$

$$\hookrightarrow +10 \sin(\omega t + \underline{50 - 90})$$

$$V_2 = +12 \sin(\omega t - 10)$$

$$V_1 = +10 \underline{\sin}(\omega t \text{ } \ominus 40) \quad \checkmark \checkmark \checkmark$$

$$V_2 = +12 \underline{\sin}(\omega t \text{ } \ominus 10) \quad \checkmark \checkmark \checkmark$$

$$\theta_1 = -40$$

$$\theta_2 = \underline{\underline{-10}}$$

$$\boxed{\theta_2} > \theta_1$$

المقادير

$$\theta_2 - \theta_1 = -40 - (-10)$$

$$-40 + 10 = \cancel{\times} \underline{\underline{30}}$$

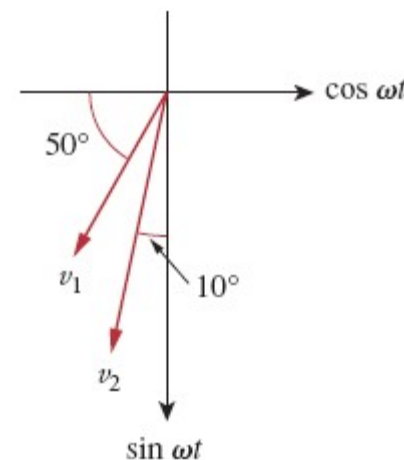
V_2 lead V_1 by 30

■ **METHOD 2** Alternatively, we may express v_1 in sine form:

$$\begin{aligned} v_1 &= -10 \cos(\omega t + 50^\circ) = 10 \sin(\omega t + 50^\circ - 90^\circ) \\ &= 10 \sin(\omega t - 40^\circ) = 10 \sin(\omega t - 10^\circ - 30^\circ) \end{aligned}$$

But $v_2 = 12 \sin(\omega t - 10^\circ)$. Comparing the two shows that v_1 lags v_2 by 30° . This is the same as saying that v_2 leads v_1 by 30° .

■ **METHOD 3** We may regard v_1 as simply $-10 \cos \omega t$ with a phase shift of $+50^\circ$. Hence, v_1 is as shown in Fig. 9.5. Similarly, v_2 is $12 \sin \omega t$ with a phase shift of -10° , as shown in Fig. 9.5. It is easy to see from Fig. 9.5 that v_2 leads v_1 by 30° , that is, $90^\circ - 50^\circ - 10^\circ$.



Sinusoid-phasor transformation

Time domain representation

Phasor domain representation

$V_m \cos(\omega t + \phi)$

$V_m \angle \phi$ ✓

$V_m \sin(\omega t + \phi)$

$V_m \angle \phi - 90^\circ$

$I_m \cos(\omega t + \theta)$

$I_m \angle \theta$

$I_m \sin(\omega t + \theta)$

$I_m \angle \theta - 90^\circ$

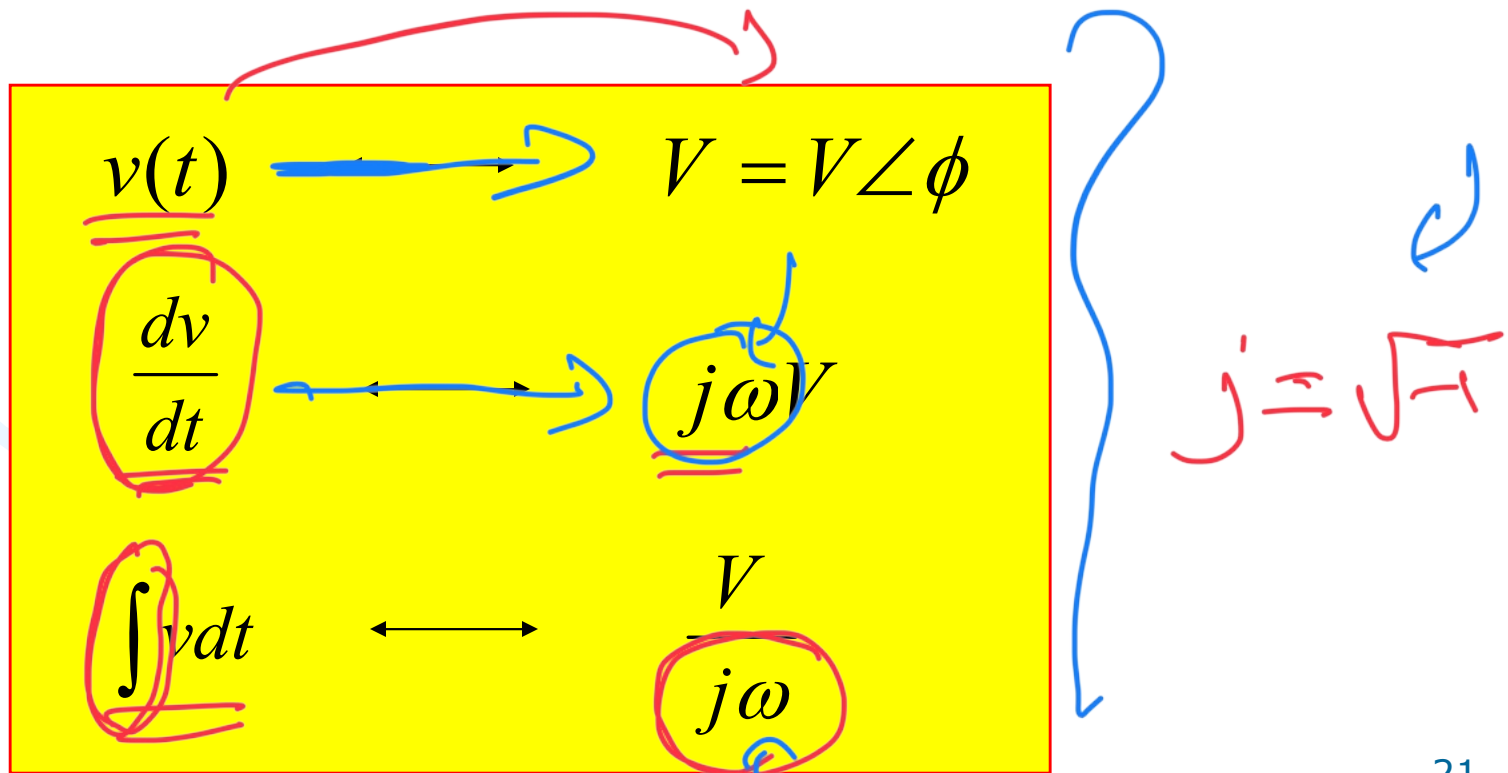
الأساس
cos

cos

لوقتك $\sin$ $\rightarrow$ $\cos$

9.3 Phasor (8)

Relationship between differential, integral operation in phasor listed as follow:



9.3 Phasor (9)

Example 8

Use phasor approach, determine the current $i(t)$ in a circuit described by the integro-differential equation.

$$4i + 8 \int i dt - 3 \frac{di}{dt} = 50 \cos(2t + 75^\circ)$$

$$\omega = 2 \text{ rad/s}$$

Answer: $i(t) = \underline{4.642 \cos(2t + 143.2^\circ)} \text{ A}$

في كتاب

2

✓

$z_j \rightarrow \text{المحاور}$

$$4\bar{I} + \frac{8\bar{I}}{j2} - (3\bar{I})(j2) =$$

So LFS

$$4\underline{\bar{I}} + -j4\underline{\bar{I}} - j6\underline{\bar{I}} = \text{So LFS}$$

$$\underline{\bar{I}} [4 - j4 - j6] = \text{So LFS}$$

$$\underline{I} \frac{4 - j10}{4 - j10} = \frac{50 \angle 75^\circ}{4 - j10}$$

amplitude

$$I = 4,642 \angle 143,199^\circ \text{ A}$$

$$i(t) = 4,642 \cos(2t + 143,199^\circ) \text{ A}$$

Solution:

We transform each term in the equation from time domain to phasor domain. Keeping Eqs. (9.27) and (9.28) in mind, we obtain the phasor form of the given equation as

$$4\mathbf{I} + \frac{8\mathbf{I}}{j\omega} - 3j\omega\mathbf{I} = 50\angle 75^\circ$$

But $\omega = 2$, so

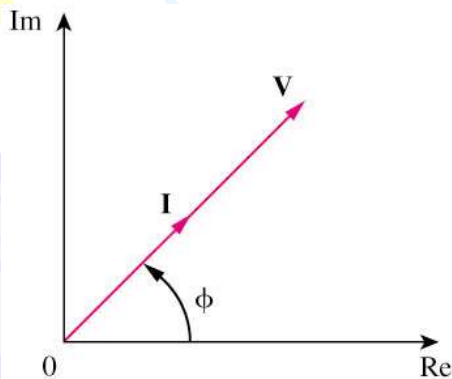
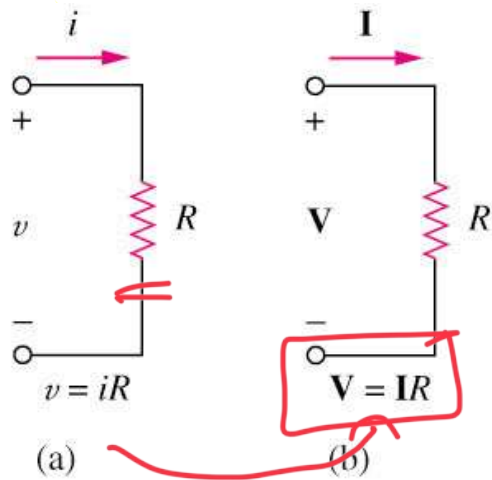
$$\mathbf{I}(4 - j4 - j6) = 50\angle 75^\circ$$
$$\mathbf{I} = \frac{50\angle 75^\circ}{4 - j10} = \frac{50\angle 75^\circ}{10.77\angle -68.2^\circ} = 4.642\angle 143.2^\circ \text{ A}$$

Converting this to the time domain,

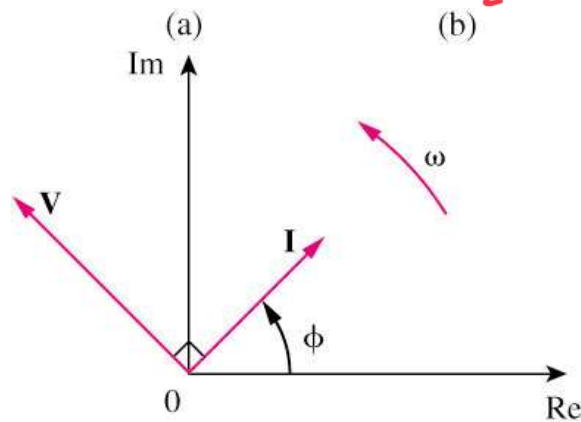
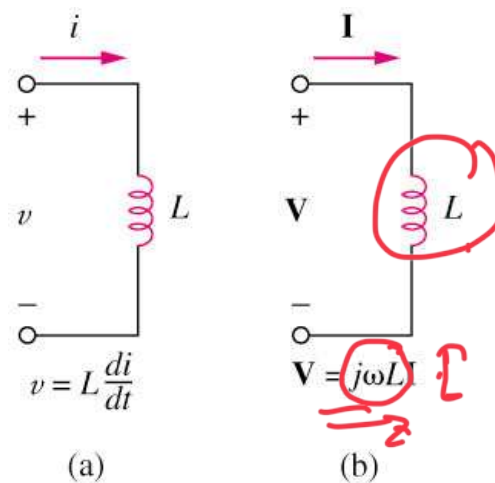
$$i(t) = 4.642 \cos(2t + 143.2^\circ) \text{ A}$$

9.4 Phasor Relationships for Circuit Elements (1)

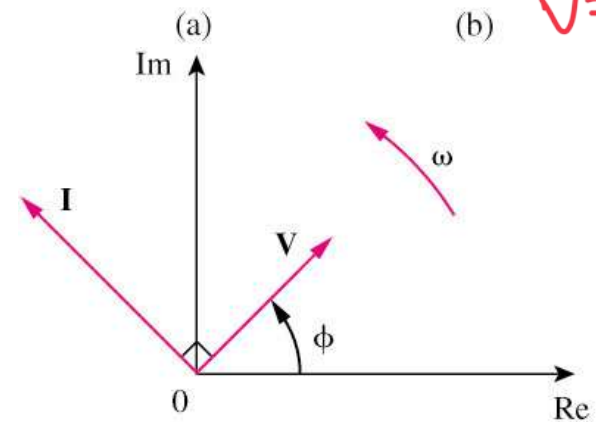
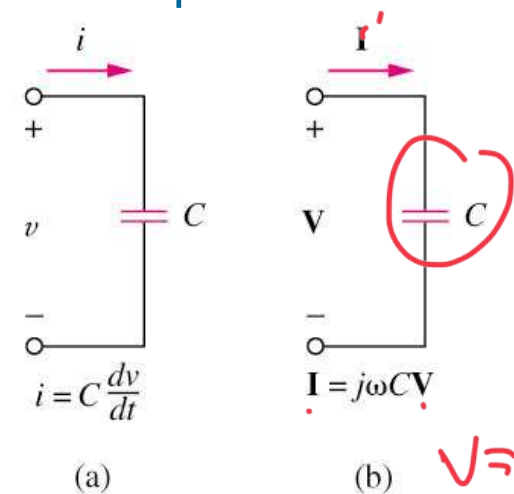
Resistor:



Inductor:



Capacitor:



$$V = \frac{I}{j\omega C}$$

Handwritten red text at the top: $\frac{1}{Z} = Y$

9.5 Impedance and Admittance (2)

Impedances and admittances of passive elements

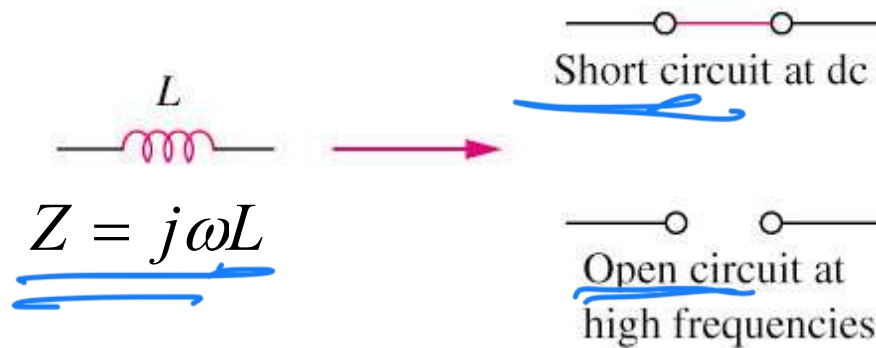
Element	<u>Impedance</u>	Admittance
R	$Z = R$	$Y = \frac{1}{R}$
L	$Z = j\omega L$	$Y = \frac{1}{j\omega L}$
C	$Z = \frac{1}{j\omega C}$	$Y = j\omega C$

Handwritten red text at the bottom: $Y = \frac{1}{Z}$

$\omega = 0 \Rightarrow DC$
 $\omega = \infty \Rightarrow AC$

scribble

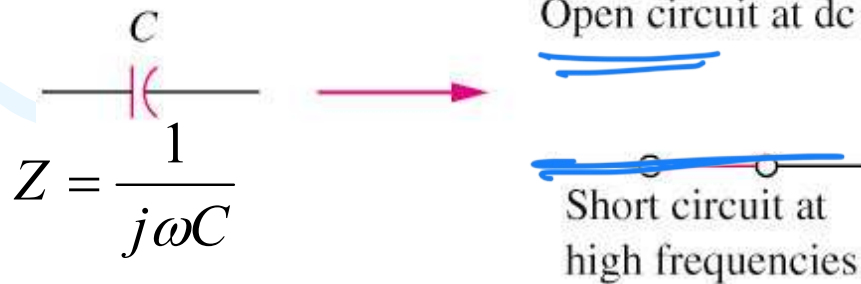
9.5 Impedance and Admittance (3)



$\omega = 0; Z = 0$

$\omega \rightarrow \infty; Z \rightarrow \infty$

(a)



$\omega = 0; Z \rightarrow \infty$

$\omega \rightarrow \infty; Z = 0$

$\frac{1}{\infty} = 0$

(b)

$R = 0 \Rightarrow \text{short}$
 $R = \infty \Rightarrow \text{open}$

9.6 Kirchhoff's Laws in the Frequency Domain (1)

- Both KVL and KCL hold in the phasor domain or more commonly called frequency domain.
- Moreover, the variables to be handled are phasors, which are complex numbers.
- All the mathematical operations involved are now in complex domain.

$$j = \sqrt{-1}$$