



# C-TOTAL IMPEDANCE FOR AC CIRCUITS

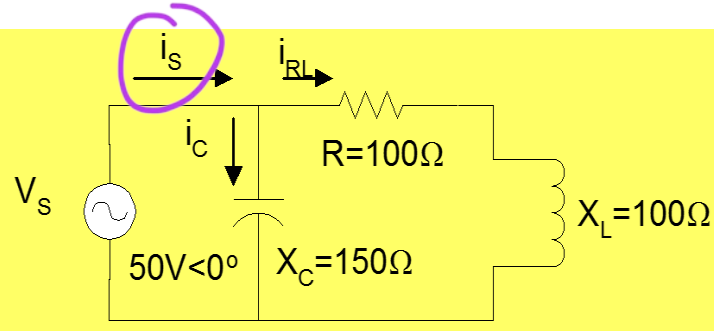
## 2-Example of Parallel circuit:

a- Find the total Impedance

b- Draw Impedance Triangle (Phasor Diagram)

c- Find  $i_s$ ,  $v_R$ ,  $v_C$ ,  $v_L$

d- Find  $v_R$ ,  $v_C$ ,  $v_L$  using Voltage Divider



✓ Solution ➤ a- The total Impedance

➤ The circuit given is already in frequency domain:

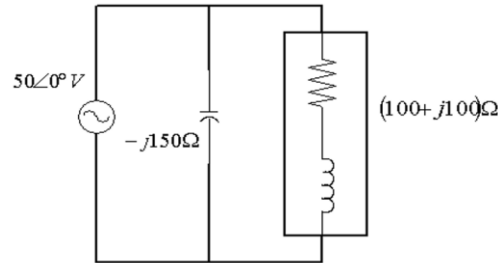
$$Z_R = 100\Omega \angle 0^\circ = 100 \Omega$$

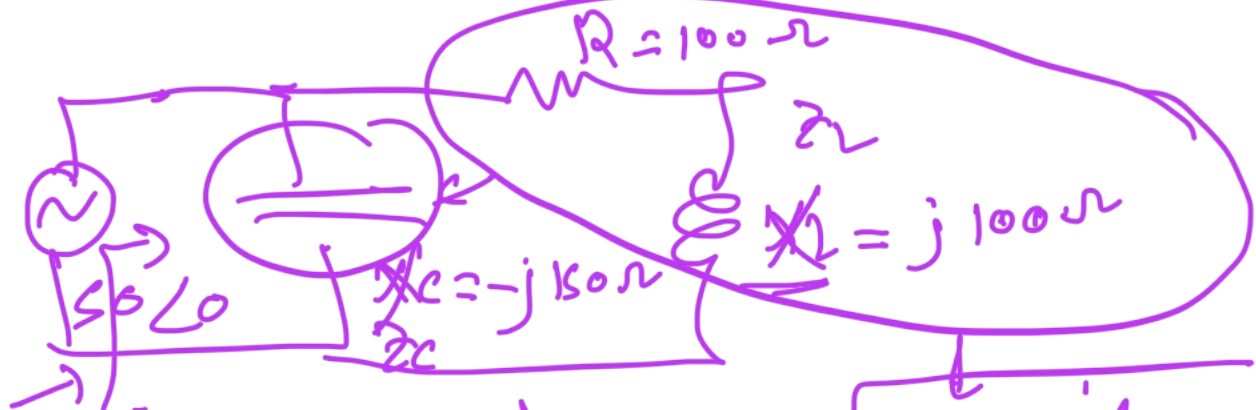
$$Z_C = 150\Omega \angle -90^\circ = -j150 \Omega$$

$$Z_L = 100\Omega \angle 90^\circ = j100 \Omega$$



➤ **Circuit simplification**





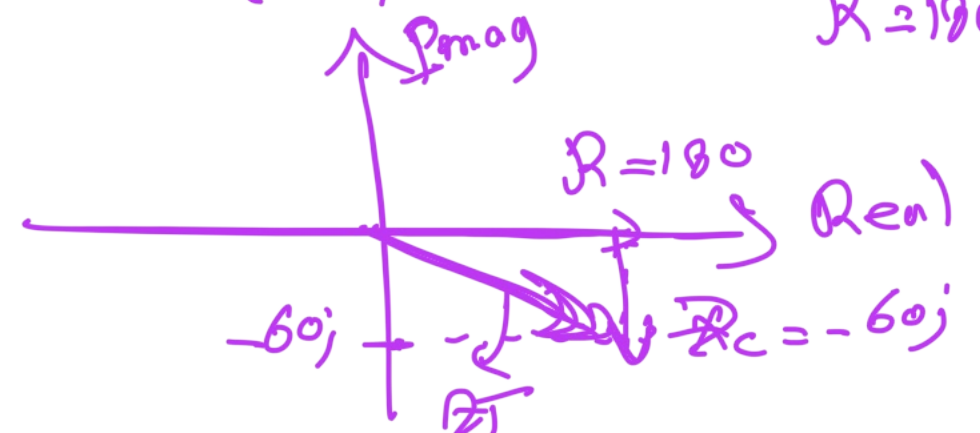
$Z_C$   
 $Z_L$  ✓  
 $Z_R$  ✓  
 ~~$X_L$~~   
 ~~$X_C$~~

$$Z_T = \frac{1}{\frac{1}{(100 + j100)} + \frac{1}{(-j150)}} = 180 - j60 \Omega$$

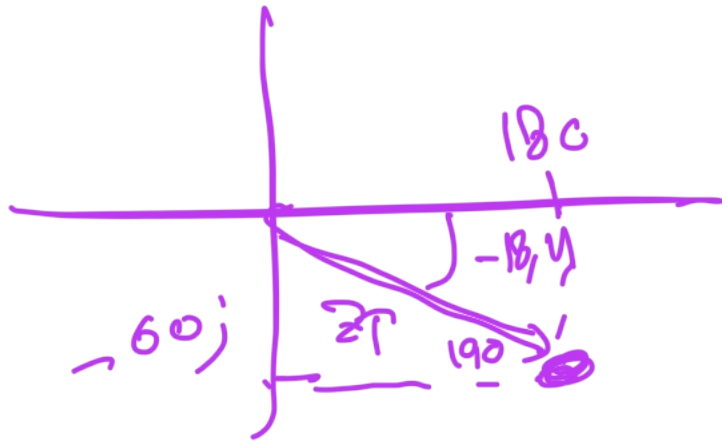
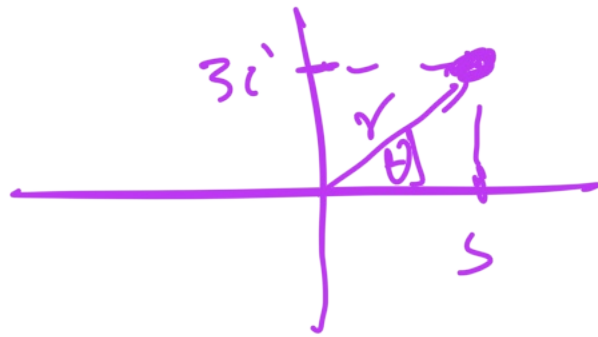
$R = 180$

$Z_L = jX_L$   
 $Z_C = -jX_C$

$X_C = 60 = \frac{1}{\omega C}$   
 $C = \frac{1}{60\omega}$



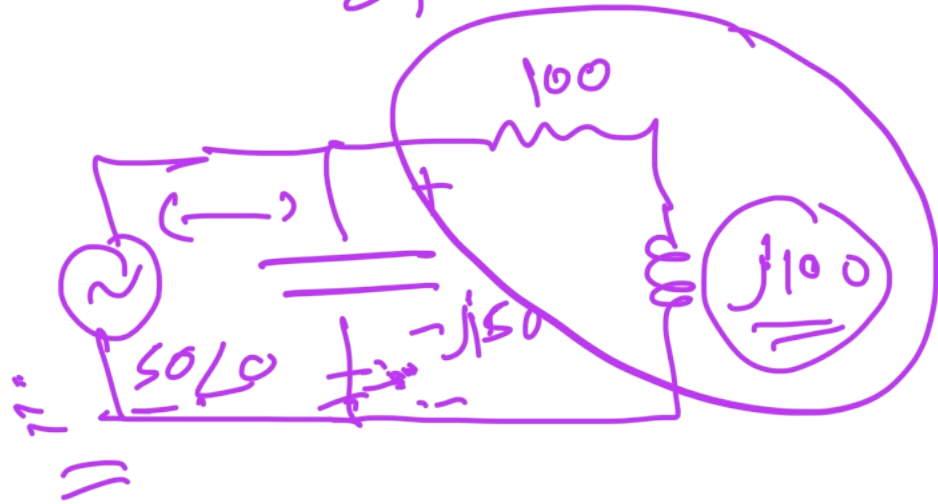
$$z = 5 + 3i$$



$$z' = 180 - j60$$

$$\underbrace{189,74}_{\theta'} \angle \underline{-18,4}$$

$$I_s = \frac{V_s}{Z_T} = \frac{50 \angle 0}{180 - j60} = 0,264 \angle 18,43 \text{ A}$$



$$V_L = V_s = 50 \angle 0$$

$$V_L = \frac{j100}{100 + j100} \times 50 \angle 0$$

$$= 25\sqrt{2} \angle 45 \text{ V}$$

$$V_R = \frac{100}{100 + j100} \times 50$$

$$= 25\sqrt{2} \angle -45 \text{ V}$$



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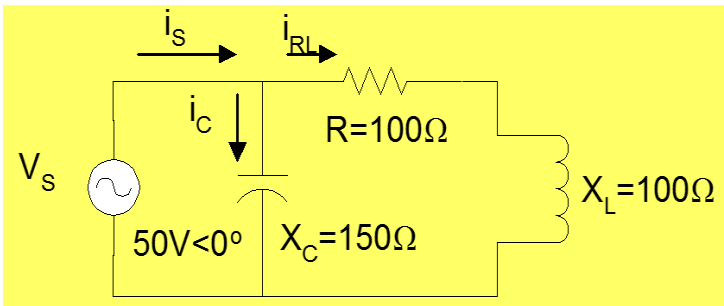
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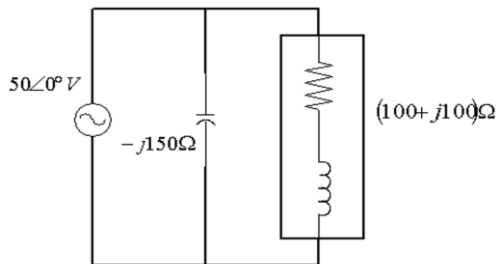
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✓ Solution ➤ a- The total Impedance

➤ The circuit given is already in frequency domain:



$$\begin{aligned} Z_{total} &= (Z_C) \parallel (Z_R + Z_L) \\ &= (150 \angle -90^\circ) \parallel [(100) + (100 \angle 90^\circ)] \\ &= (150 \angle -90^\circ) \parallel (100 + j100) \\ &= (150 \angle -90^\circ) \parallel (141.42 \angle 45^\circ) \end{aligned}$$

$$\begin{aligned} \therefore Z_{total} &= [Y_{total}]^{-1} \\ &= \left[ \frac{1}{150 \angle -90^\circ} + \frac{1}{141.42 \angle 45^\circ} \right]^{-1} \end{aligned}$$

$$\therefore Z_{total} = \left( \frac{1}{150} (\cos(90^\circ) + j \sin(90^\circ)) + \frac{1}{141.42} (\cos(-45^\circ) + j \sin(-45^\circ)) \right)^{-1}$$

$$= (0.005 + j0.0016)^{-1}$$

$$= \left( \sqrt{0.005^2 + 0.0016^2} \angle \tan^{-1}\left(\frac{0.0016}{0.005}\right) \right)^{-1}$$

$$= (0.00524 \angle 17.744^\circ)^{-1} \approx 190 \angle -18^\circ$$

$$\therefore Z_{total} \approx 180 - j60 \Omega$$



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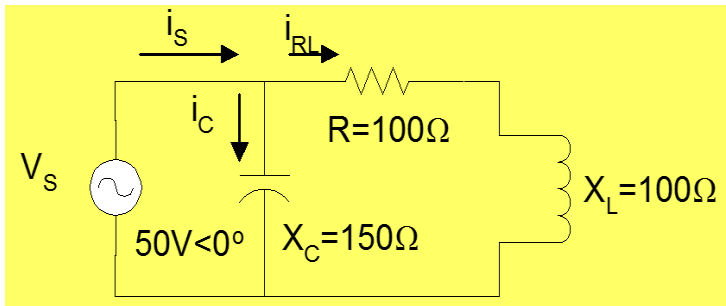
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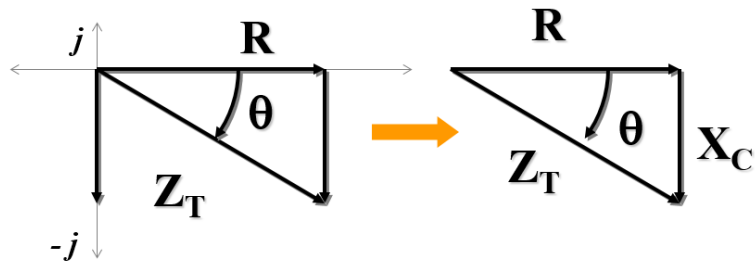
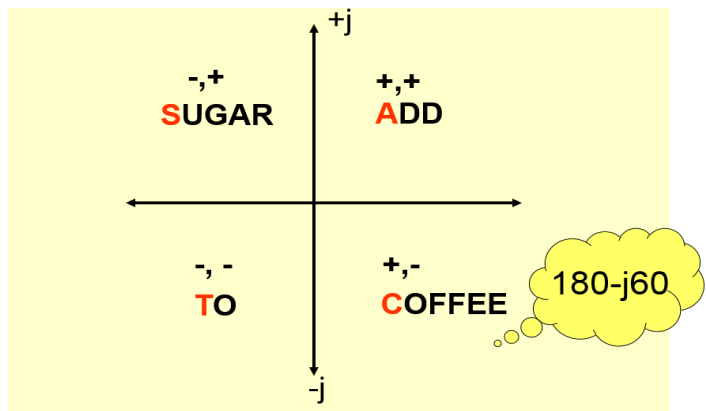
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✓ Solution ➤ b- The Impedance Triangle (Phasor Diagram)

$$\therefore Z_{total} \simeq 180 - j60 \Omega$$





# C-TOTAL IMPEDANCE FOR AC CIRCUITS

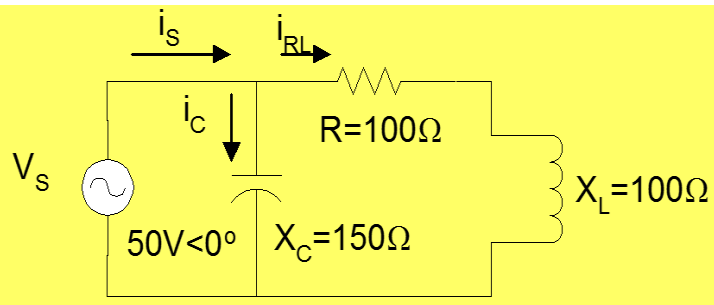
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✓ Solution      ➤ c-  $i_s$ ,  $v_R$ ,  $v_C$ ,  $v_L$ :

$$\therefore Z_{total} \simeq 180 - j60 \Omega$$

$$i_s = \frac{v_s}{Z_T} = \frac{50 \angle 0^\circ}{190 \angle -18^\circ} = 263 \angle 18^\circ \text{ mA}$$

$$v_S = v_C = (v_R + v_L) = 50 \angle 0^\circ \text{ V}$$

$$\Rightarrow i_C = \frac{v_S}{Z_C} = \frac{50 \angle 0^\circ}{150 \angle -90^\circ} = 333 \angle 90^\circ \text{ mA}$$

$$\Rightarrow i_{RL} = \frac{v_S}{Z_R + Z_L} = \frac{50 \angle 0^\circ}{100 + (100 \angle -90^\circ)} = 353 \angle -45^\circ \text{ mA}$$



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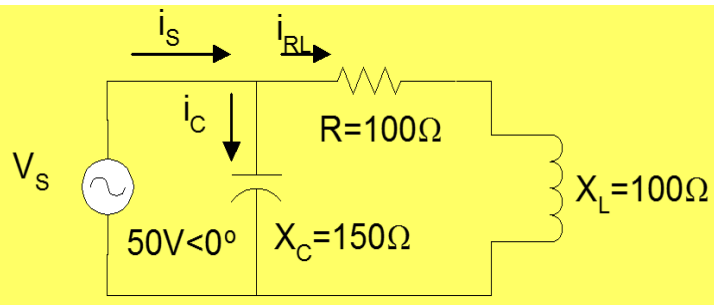
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✓ Solution ➤ d- Find  $v_R$ ,  $v_C$ ,  $v_L$  using Current Divider

$$i_c = \frac{Z_T}{Z_C} i_S = \frac{(190 \angle -18^\circ)}{(150 \angle -90^\circ)} \cdot (263 \angle 18^\circ \text{ mA})$$
$$= 333 \angle 90^\circ \text{ mA}$$

$$i_{RL} = \frac{Z_T}{Z_R + Z_L} i_S = \frac{(190 \angle -18^\circ)}{(100)(100 \angle 90^\circ)} \cdot (263 \angle 18^\circ \text{ mA})$$
$$= 353 \angle -45^\circ \text{ mA}$$



# D-Impedance as a Function of Frequency

➤ The Impedance  $Z$  of a circuit is a function of the frequency.

Element	Impedance
$L$	$Z = j\omega L$
$C$	$Z = \frac{1}{j\omega C}$

➤ Inductor is **SHORT CIRCUIT** at DC and **OPEN CIRCUIT** at high frequencies.

Capacitor is **OPEN CIRCUIT** at DC and **SHORT CIRCUIT** at high frequencies.

$$Z_L = j\omega L$$

$$Z_L \rightarrow 0 \quad \omega \rightarrow 0 \quad (\text{Short at DC})$$

$$Z_L \rightarrow \infty \quad \omega \rightarrow \infty \quad (\text{Open as } \omega \rightarrow \infty)$$

$$Z_C = \frac{1}{j\omega C}$$

$$Z_C \rightarrow \infty \quad \omega \rightarrow 0 \quad (\text{Open at DC})$$

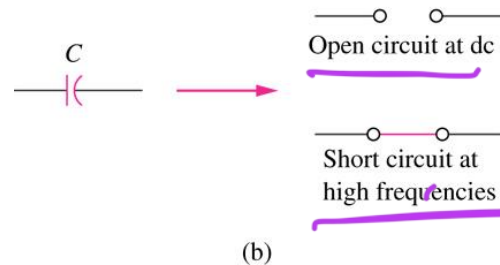
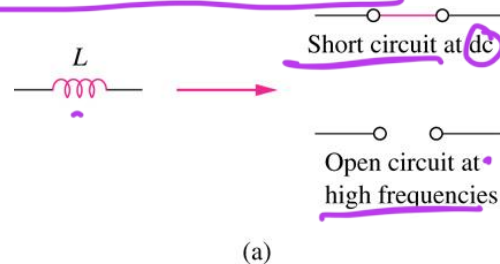
$$Z_C \rightarrow 0 \quad \omega \rightarrow \infty \quad (\text{Short as } \omega \rightarrow \infty)$$

short

$\omega \rightarrow 0$

$\omega \rightarrow \infty$

$\omega = \infty$





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# AC Analysis Techniques

# AC Analysis Techniques

- Both **KVL** and **KCL** are hold in Phasor domain or more commonly called frequency domain.
- Moreover, the variables to be handled are phasors, which are complex numbers.
- All the mathematical operations involved are now in complex domain.
- **All DC circuit analysis principles apply to AC circuits.**
  - Voltage Division
  - Current Division
  - Circuit Reduction
  - Impedance equivalence
  - Y-Delta Transformation

# Example 22: Voltage Divider Rule

➤ Calculate the  $v_0$  in the given circuit

➤ Solution:

In the frequency domain,

the voltage source is  $V_s = 10\angle 75^\circ$

the 0.5-H inductor is  $j\omega L = j(10)(0.5) = j5$

the  $\frac{1}{20}$ -F capacitor is  $\frac{1}{j\omega C} = \frac{1}{j(10)(1/20)} = -j2$

Let  $Z_1 =$  impedance of the 0.5-H inductor in parallel with the 10- $\Omega$  resistor

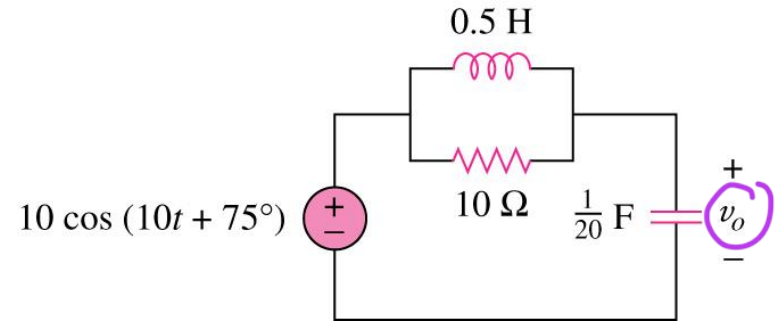
and  $Z_2 =$  impedance of the (1/20)-F capacitor

$$Z_1 = 10 \parallel j5 = \frac{(10)(j5)}{10 + j5} = 2 + j4 \quad \text{and} \quad Z_2 = -j2$$

$$V_o = Z_2 / (Z_1 + Z_2) V_s$$

$$V_o = \frac{-j2}{2 + j4 - j2} (10\angle 75^\circ) = \frac{-j(10\angle 75^\circ)}{1 + j} = \frac{10\angle (75^\circ - 90^\circ)}{\sqrt{2}\angle 45^\circ} = 7.071\angle -60^\circ$$

$$v_o(t) = \underline{7.071 \cos(10t - 60^\circ) \text{ V}}$$



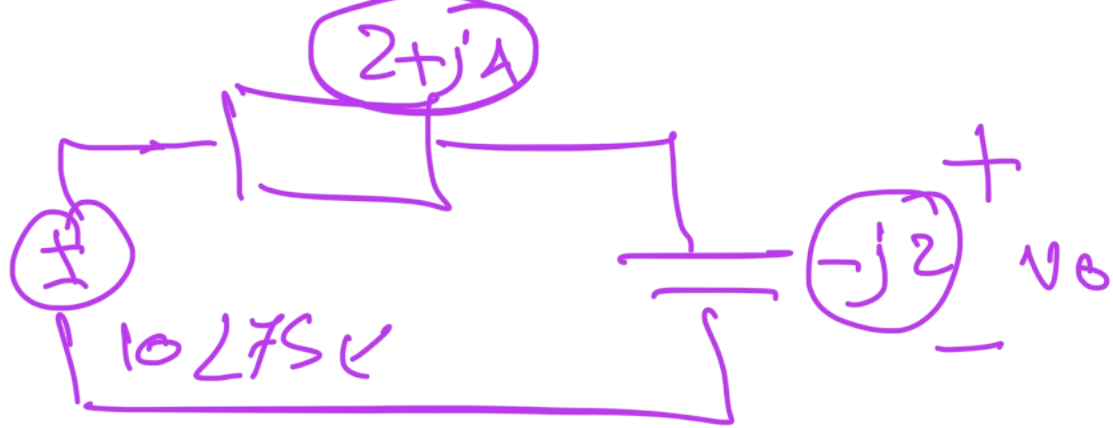


$10 \cos(10t + \pi/5)$   
 $\omega = 10$

$10 \angle \pi/5$

$Z = \frac{1}{\frac{1}{10} + \frac{1}{j5}}$   
 $= 2 + j4$

$Z_R = R$   
 $Z_L = j\omega L$   
 $Z_C = \frac{1}{j\omega C}$



Voltage divider

$$V_0 = \frac{-j2}{-j2 + 2 + j4} \times 10\angle 75^\circ = \boxed{5\sqrt{2} \angle -60^\circ} \text{ V}$$

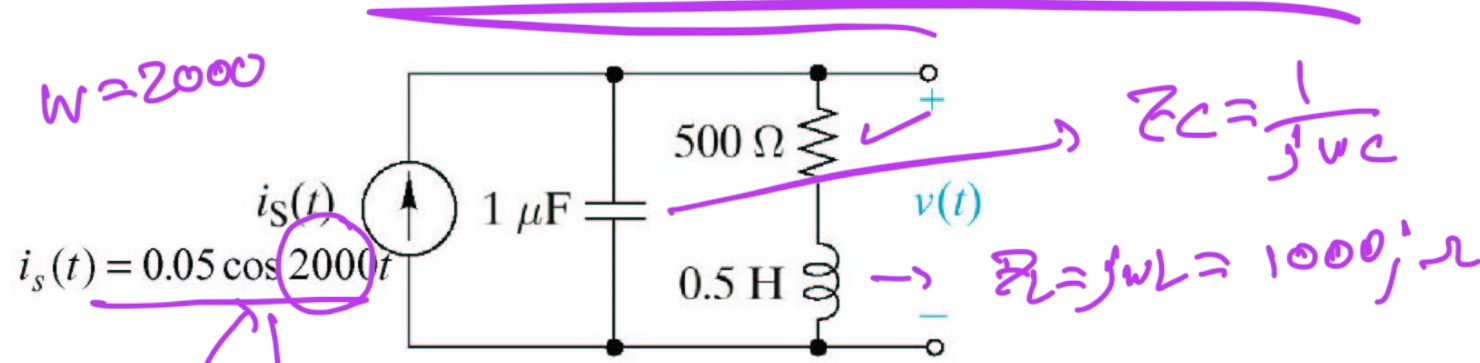
$\uparrow$  7,07 A

$$\boxed{V_0(t) = 5\sqrt{2} \cos(10t - 60^\circ)} \text{ V}$$

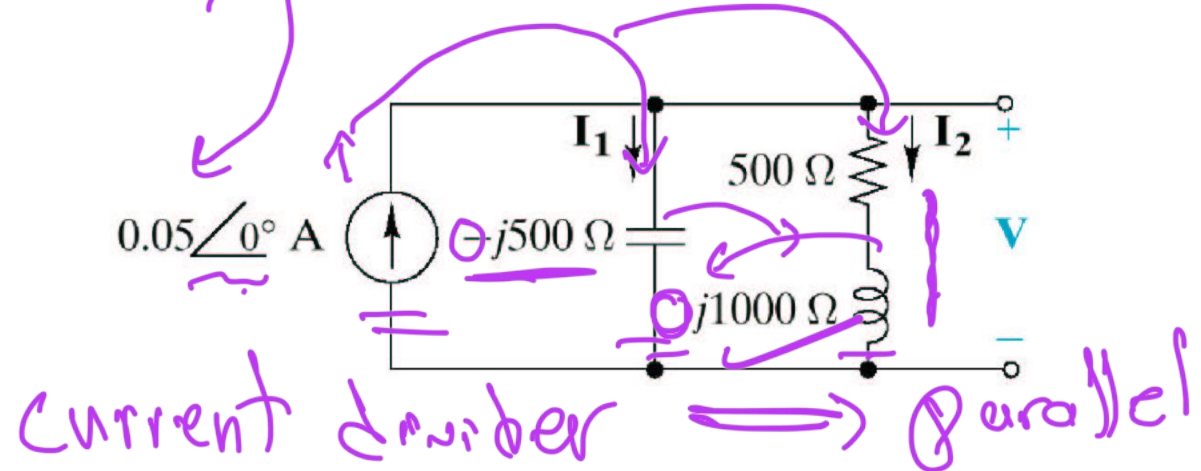
Steady state  
Time domain

# Example 23: Current Divider Rule

Find  $I_1$  and  $I_2$  in phasor domain, then convert them into time domain



Solution:



$$I_1 = \frac{500 + j1000}{500 + j1000 - j500} \times 0,05 = \underline{0,079} \angle 18,43^\circ \text{ A}$$

$$I_2 = \frac{-j500}{-j500 + 500 + j1000} \times 0,05 = \underline{0,035} \angle -13,5^\circ \text{ A}$$

$$I_1 = 79 \cos(2000t + 18,43^\circ) \text{ mA}$$

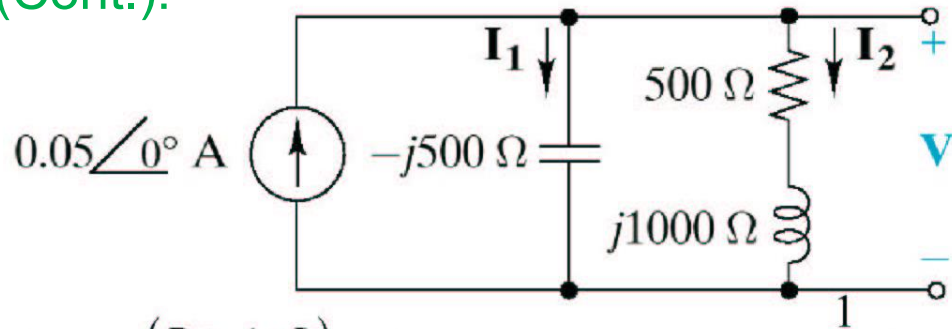
$$I_2 = 35 \cos(2000t - 13,5^\circ) \text{ mA}$$

steady

state

# Example 23: Current Divider Rule

Solution (Cont.):



$$\hat{I}_1 = \frac{(R + j\omega L)}{R + j\omega L + \frac{1}{j\omega C}} \hat{I}$$

$$\hat{I}_1 = \frac{(500 + j1000)}{500 + j1000 - j500} 0.05 \angle 0^\circ$$

$$I_1 = 0.079 \angle 18.4^\circ \text{ A}$$

$$\hat{I}_2 = \frac{\frac{1}{j\omega C}}{R + j\omega L + \frac{1}{j\omega C}} \hat{I}$$

$$\hat{I}_2 = \frac{-j500}{500 + j1000 - j500} 0.05 \angle 0^\circ$$

$$I_2 = 0.03535 \angle -135^\circ \text{ A}$$

$$i_1(t) = 79 \cos(2000t + 18.4^\circ) \text{ mA}$$

$$i_2(t) = 35.35 \cos(2000t - 135^\circ) \text{ mA}$$

# Delta-Wye ( $\Delta - Y$ ) Transformation

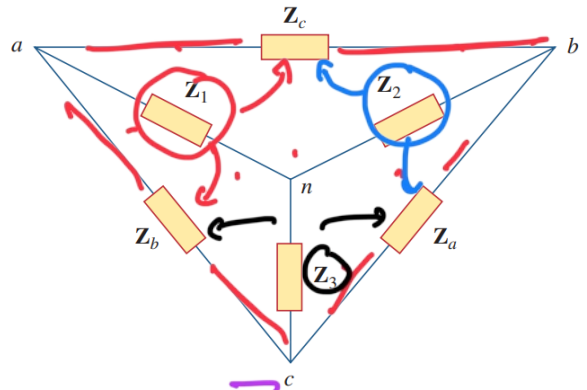
تحويل

Y- $\Delta$  Conversion:

$$Z_a = \frac{Z_1 Z_2 + Z_2 Z_3 + Z_3 Z_1}{Z_1}$$

$$Z_b = \frac{Z_1 Z_2 + Z_2 Z_3 + Z_3 Z_1}{Z_2}$$

$$Z_c = \frac{Z_1 Z_2 + Z_2 Z_3 + Z_3 Z_1}{Z_3}$$



تحويل

$\Delta$ -Y Conversion:

$$Z_1 = \frac{Z_b Z_c}{Z_a + Z_b + Z_c}$$

$$Z_2 = \frac{Z_c Z_a}{Z_a + Z_b + Z_c}$$

$$Z_3 = \frac{Z_a Z_b}{Z_a + Z_b + Z_c}$$

# Example 24- ( $\Delta - Y$ ) Transformation

Find current  $\mathbf{I}$  in the circuit

Solution:

$$\mathbf{Z}_{an} = \frac{j4(2 - j4)}{j4 + 2 - j4 + 8} = \frac{4(4 + j2)}{10} = (1.6 + j0.8) \Omega$$

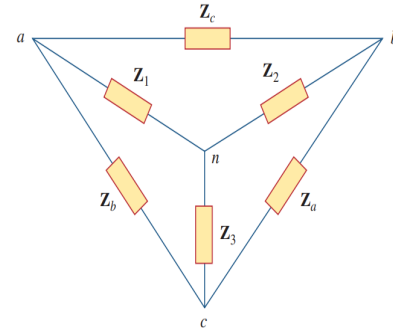
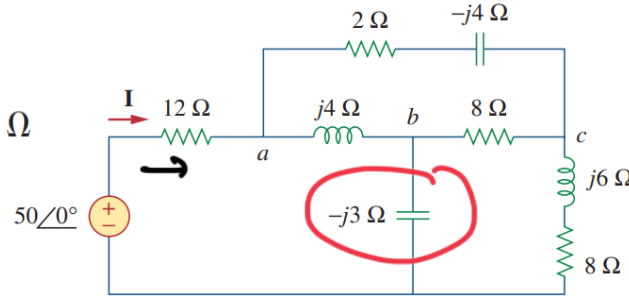
$$\mathbf{Z}_{bn} = \frac{j4(8)}{10} = j3.2 \Omega, \quad \mathbf{Z}_{cn} = \frac{8(2 - j4)}{10} = (1.6 - j3.2) \Omega$$

The total impedance at the source terminals is

$$\begin{aligned} \mathbf{Z} &= 12 + \mathbf{Z}_{an} + (\mathbf{Z}_{bn} - j3) \parallel (\mathbf{Z}_{cn} + j6 + 8) \\ &= 12 + 1.6 + j0.8 + (j0.2) \parallel (9.6 + j2.8) \\ &= 13.6 + j0.8 + \frac{j0.2(9.6 + j2.8)}{9.6 + j3} \\ &= 13.6 + j1 = 13.64 \angle 4.204^\circ \end{aligned}$$

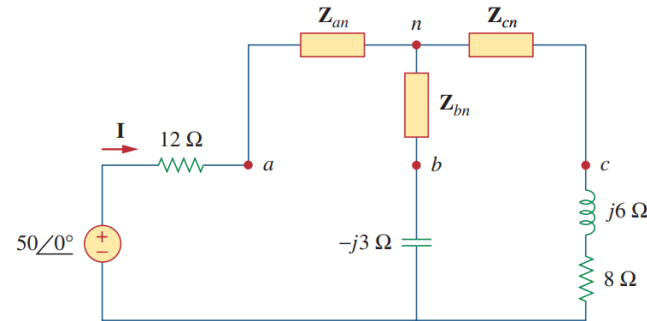
The desired current is

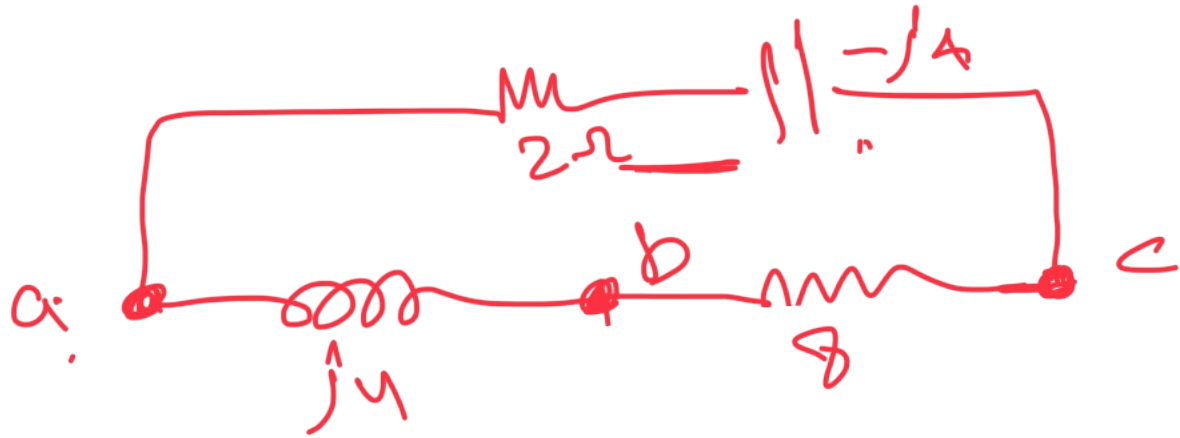
$$\mathbf{I} = \frac{\mathbf{V}}{\mathbf{Z}} = \frac{50 \angle 0^\circ}{13.64 \angle 4.204^\circ} = 3.666 \angle -4.204^\circ \text{ A}$$



$\Delta$ -Y Conversion:

$$\begin{aligned} \mathbf{Z}_1 &= \frac{\mathbf{Z}_b \mathbf{Z}_c}{\mathbf{Z}_a + \mathbf{Z}_b + \mathbf{Z}_c} \\ \mathbf{Z}_2 &= \frac{\mathbf{Z}_c \mathbf{Z}_a}{\mathbf{Z}_a + \mathbf{Z}_b + \mathbf{Z}_c} \\ \mathbf{Z}_3 &= \frac{\mathbf{Z}_a \mathbf{Z}_b}{\mathbf{Z}_a + \mathbf{Z}_b + \mathbf{Z}_c} \end{aligned}$$

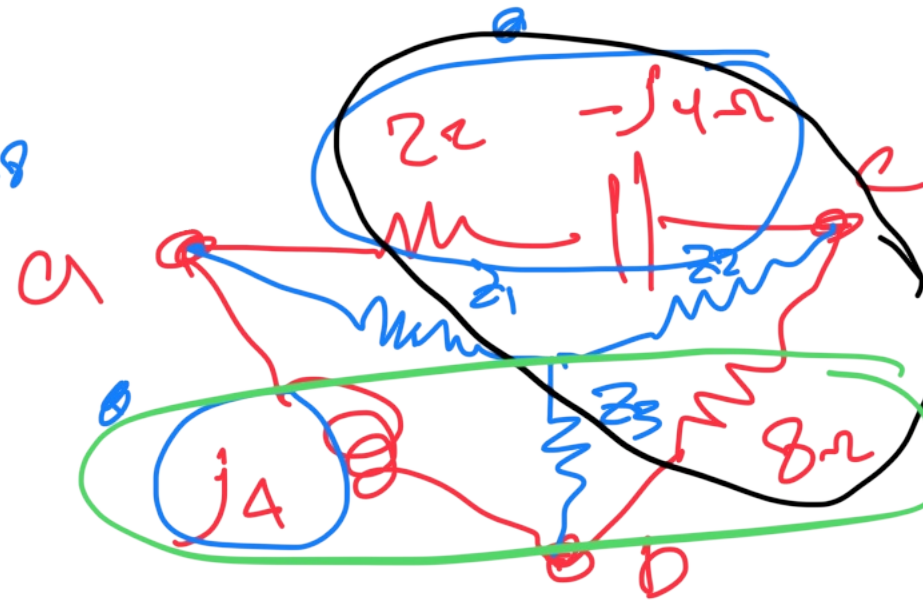


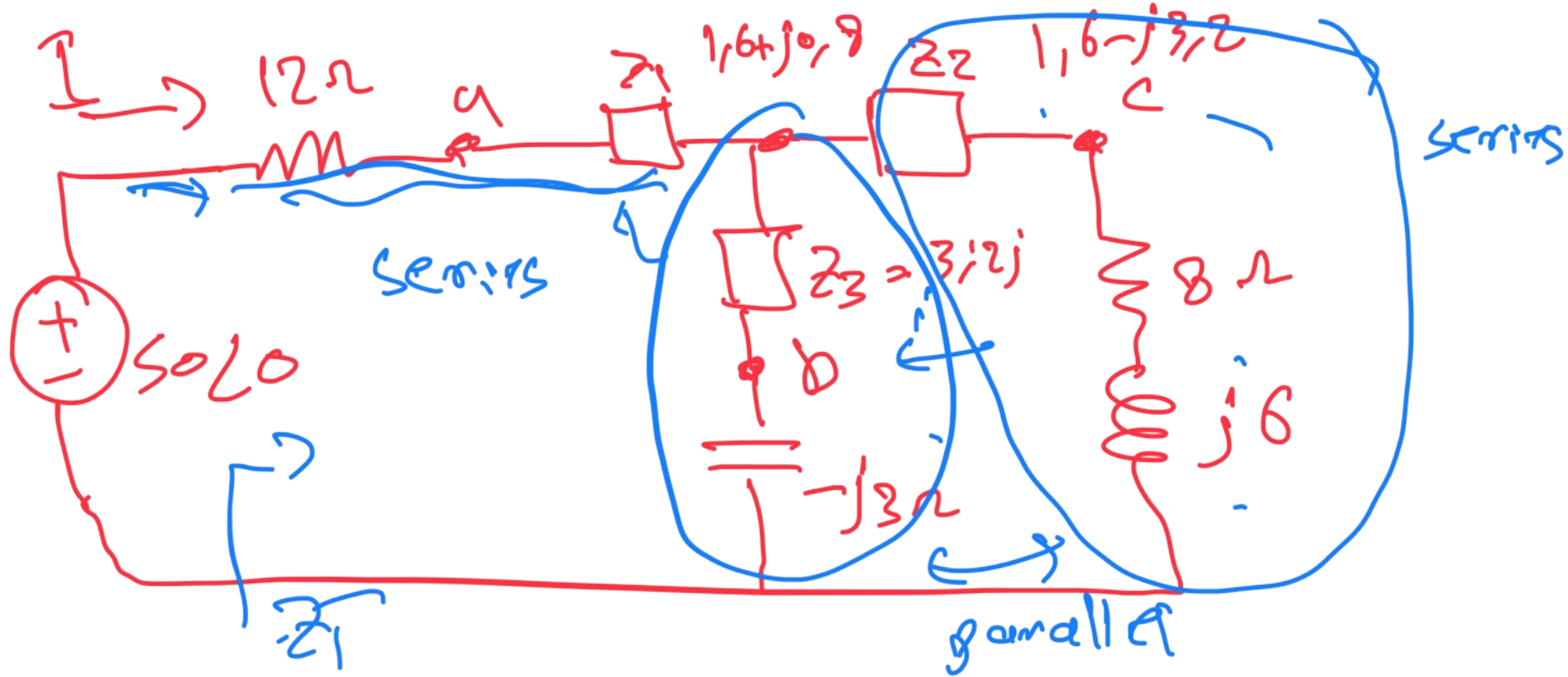


قدیم  
 $\Delta \rightarrow Y$

$$Z_1 = \frac{(j4)(2 - j4)}{j4 + 2 - j4 + 8} = 1,6 + j0,8$$

$$Z_2 = \frac{8(2 - j4)}{10} = 1,6 - j3,2$$





$$Z_T = 12 + 1,6 + j0,8 +$$


$$\frac{1}{(3j - j3)} + \frac{1}{8 + j6 + 1,6 - j3,2}$$

$$Z_T = 13,64 \angle 4,2^\circ \Omega$$

$$I = \frac{V}{Z_T} = \frac{50 \angle 20^\circ}{13,64 \angle 4,2^\circ} = 3,668 \angle -4,2^\circ \text{ A}$$

$$Z_3 = \frac{8 \times j^4}{10} = 3,2j$$

## Frequency Domain Analysis (Using Phasor)



Nodal Analysis  
Mesh Analysis  
Superposition,  
Source transformation  
Thevenin theorem, and  
Norton theorem

# Example 25- Nodal Analysis

Find  $i_x$  in the following circuit using nodal analysis

## Solution

We first convert the circuit to the frequency domain:

$$20 \cos 4t \Rightarrow 20 \angle 0^\circ, \quad \omega = 4 \text{ rad/s}$$

$$1 \text{ H} \Rightarrow j\omega L = j4$$

$$0.5 \text{ H} \Rightarrow j\omega L = j2$$

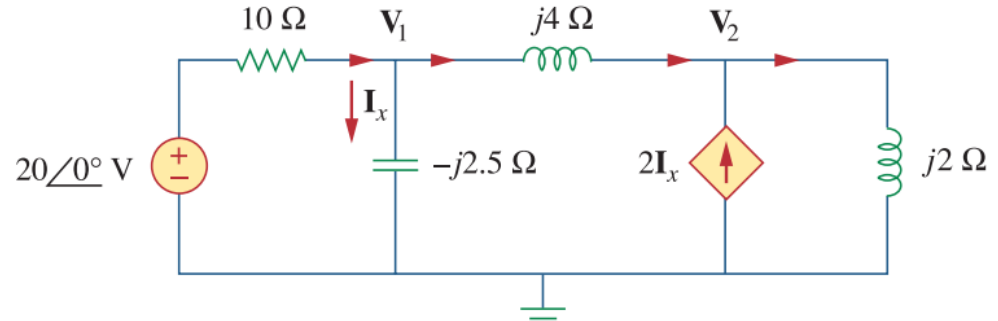
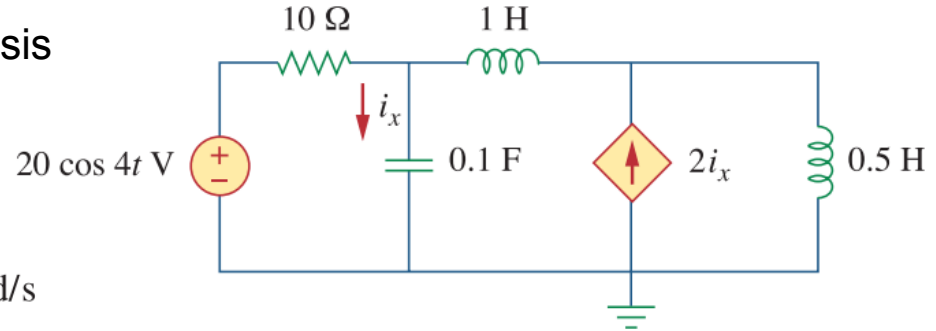
$$0.1 \text{ F} \Rightarrow \frac{1}{j\omega C} = -j2.5$$

Applying KCL at node 1,

$$\frac{20 - \mathbf{V}_1}{10} = \frac{\mathbf{V}_1}{-j2.5} + \frac{\mathbf{V}_1 - \mathbf{V}_2}{j4}$$

or

$$(1 + j1.5)\mathbf{V}_1 + j2.5\mathbf{V}_2 = 20$$



# Example 25- Nodal Analysis

Find  $i_x$  in the following circuit using nodal analysis

Solution (continue)

At node 2,

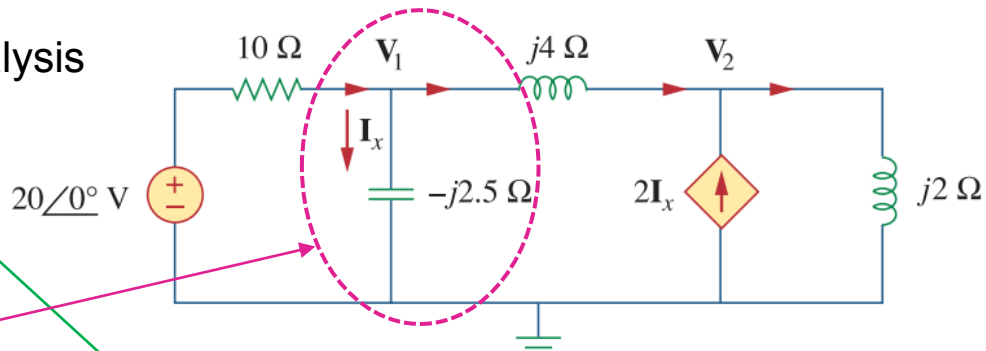
$$(1 + j1.5)V_1 + j2.5V_2 = 20$$

$$2\mathbf{I}_x + \frac{V_1 - V_2}{j4} = \frac{V_2}{j2}$$

But  $\mathbf{I}_x = V_1 / -j2.5$ . Substituting this gives

$$\frac{2V_1}{-j2.5} + \frac{V_1 - V_2}{j4} = \frac{V_2}{j2}$$

$$11V_1 + 15V_2 = 0$$



$$\begin{bmatrix} 1 + j1.5 & j2.5 \\ 11 & 15 \end{bmatrix} \begin{bmatrix} V_1 \\ V_2 \end{bmatrix} = \begin{bmatrix} 20 \\ 0 \end{bmatrix}$$

$$V_1 = 18.97 / \underline{18.43^\circ} \text{ V}$$

$$V_2 = 13.91 / \underline{198.3^\circ} \text{ V}$$

The current  $\mathbf{I}_x$  is given by

$$\mathbf{I}_x = \frac{V_1}{-j2.5} = \frac{18.97 / \underline{18.43^\circ}}{2.5 / \underline{-90^\circ}} = 7.59 / \underline{108.4^\circ} \text{ A}$$

$$\longrightarrow i_x = 7.59 \cos(4t + 108.4^\circ) \text{ A}$$

# Example 26- Mesh Analysis

Find  $I_o$  in the following circuit using mesh analysis

## Solution

Applying KVL to mesh 1, we obtain

$$\begin{aligned} &\text{➤ } 8I_1 + j10(I_1 - I_3) - 2j(I_1 - I_2) = 0 \\ (8 + j10 - j2)I_1 - (-j2)I_2 - j10I_3 &= 0 \end{aligned}$$

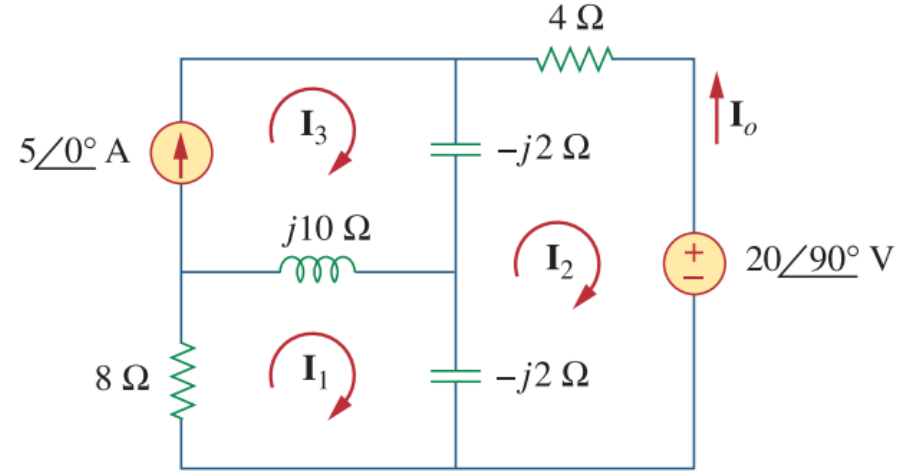
$$\text{For mesh 2, } \text{➤ } 20\angle 90^\circ + 4I_2 - j2(I_2 - I_3) - j2(I_2 - I_1) = 0$$

$$(4 - j2 - j2)I_2 - (-j2)I_1 - (-j2)I_3 + 20\angle 90^\circ = 0$$

For mesh 3,  $I_3 = 5$

$$\begin{bmatrix} 8 + j8 & j2 \\ j2 & 4 - j4 \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \end{bmatrix} = \begin{bmatrix} j50 \\ -j30 \end{bmatrix}$$

$$I_2 = \frac{\Delta_2}{\Delta} = \frac{416.17\angle -35.22^\circ}{68} = 6.12\angle -35.22^\circ \text{ A} \longrightarrow I_o = -I_2 = 6.12\angle 144.78^\circ \text{ A}$$

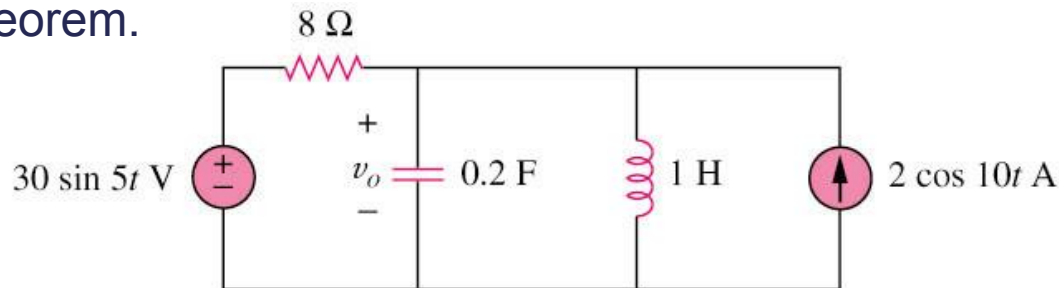


# Superposition Theorem

When a circuit has sources operating at different frequencies,

- The separate phasor circuit for each frequency must be solved independently, and
- The total response is the sum of time-domain responses of all the individual phasor circuits.

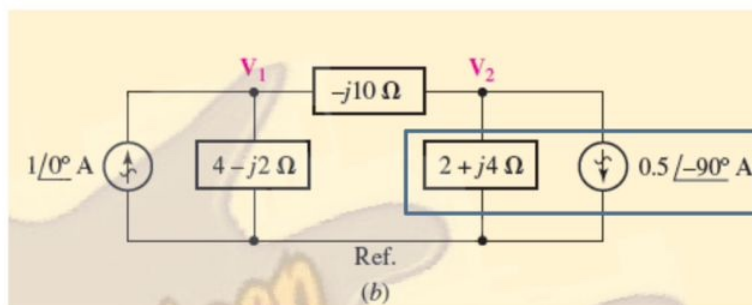
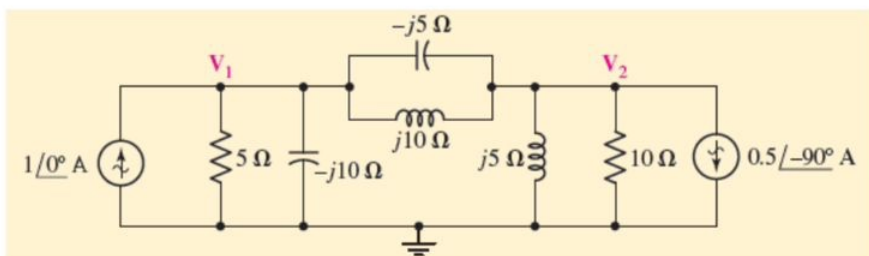
**Example** Calculate  $v_o$  in the circuit of figure shown below using the superposition theorem.



**Answer:**  $V_o = 4.631 \sin(5t - 81.12) + 1.051 \cos(10t - 86.24) \text{ V}$

# Example 27: Superposition Theorem

- Calculate  $v_1$  in the circuit of figure shown below using the superposition theorem.



**Answer:**

First we redraw the circuit as Fig. *b* where each pair of parallel impedances is replaced by a single equivalent impedance. That is,  $5 \parallel -j10 \Omega$  is  $4 - j2 \Omega$ ;  $j10 \parallel -j5 \Omega$  is  $-j10 \Omega$ ; and  $10 \parallel j5$  is equal to  $2 + j4 \Omega$ .

To find  $V_1$ , we first activate only the left source and find the partial response,  $V_{1L}$ . The  $1/0^\circ$  source is in parallel with an impedance of

$$(4 - j2) \parallel (-j10 + 2 + j4) \text{ and we have: } V_{1L} = (4 - j2) * i_1$$

so that

$$\begin{aligned} V_{1L} &= 1/0^\circ \frac{(4 - j2)(-j10 + 2 + j4)}{4 - j2 - j10 + 2 + j4} \\ &= \frac{-4 - j28}{6 - j8} = 2 - j2 \text{ V} \end{aligned}$$

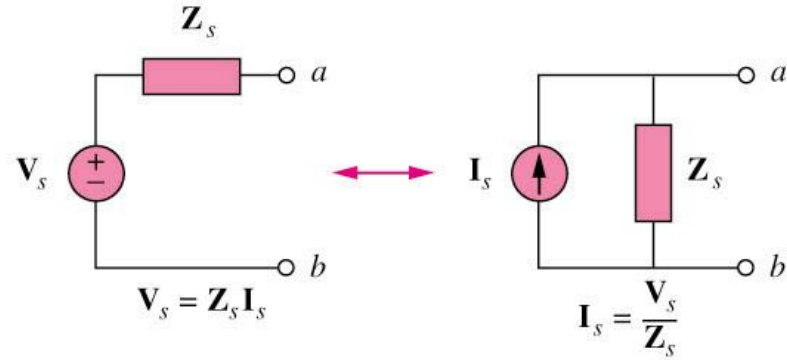
With only the right source active, current division and Ohm's law yield :  $V_{1R} = (4 - j2) * i_1'$

$$V_{1R} = (-0.5/-90^\circ) \left( \frac{2 + j4}{4 - j2 - j10 + 2 + j4} \right) (4 - j2) = -1 \text{ V}$$

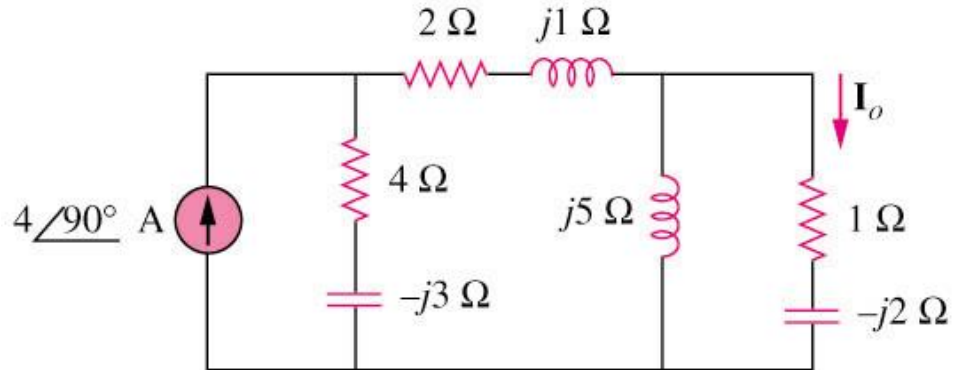
Summing, then,

$$V_1 = V_{1L} + V_{1R} = 2 - j2 - 1 = 1 - j2 \text{ V}$$

# Source Transformation



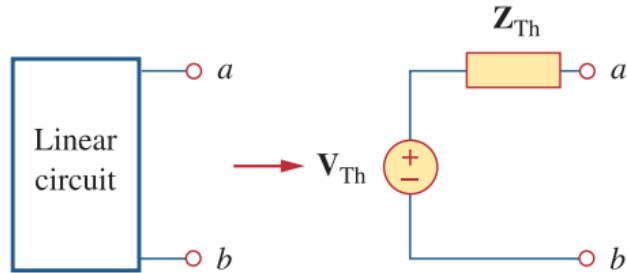
**Example** Find  $I_o$  in the circuit of figure below using the concept of source transformation.



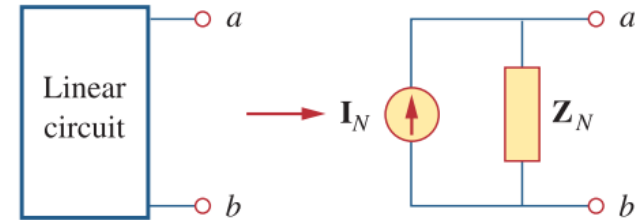
**Answer:**  $I_o = \underline{3.288 + j99.46 \text{ A}}$

# Thevenin and Norton Theorems in AC circuits

- Thevenin's and Norton theorems are applied to ac circuits in the same way as they are to DC circuits
- The **ONLY additional effort** is the need to manipulate **complex numbers**.



Thevenin equivalent.

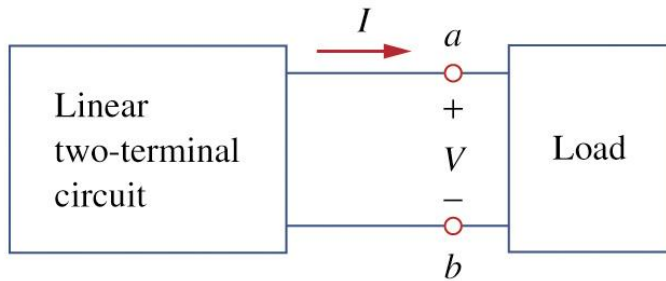


Norton equivalent.

Where;  $V_{Th} = V_{ab-OC}$  is the **open circuit voltage (PHASOR)** between terminals a-b,  $I_N = I_{ab-SC}$  is the **short circuit current (PHASOR)** through the terminals a-b, and  $Z_{Th}$  is the **input or equivalent IMPEDANCE** at the terminals **when the independent source are turn off**.

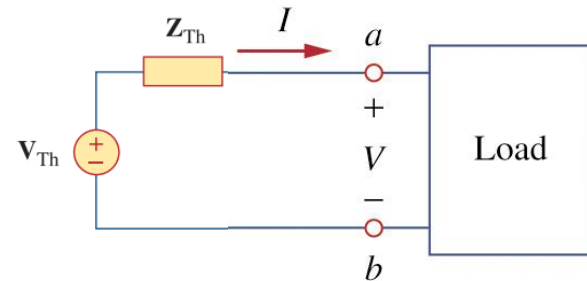
# How to Find Thevenin Equivalent Circuit in AC circuits

- ❖ First, **Open the circuit** (remove the load) at the **points** of interest **a-b**
- 1-  $V_{th}$  = Open circuit voltage (keep all sources intact). Note:  $V_{th}$  is Phasor
- 2-  $Z_{th}$  = Open circuit **equivalent Impedance** appears at terminals **a-b** while all **independent sources=0** (voltage source=SC, current source=OC).



Original Circuit

≡



Thevenin equivalent circuit

# Example 10.8- Thevenin Equivalent Circuit

Obtain the Thevenin equivalent at terminals a-b of the following circuit.

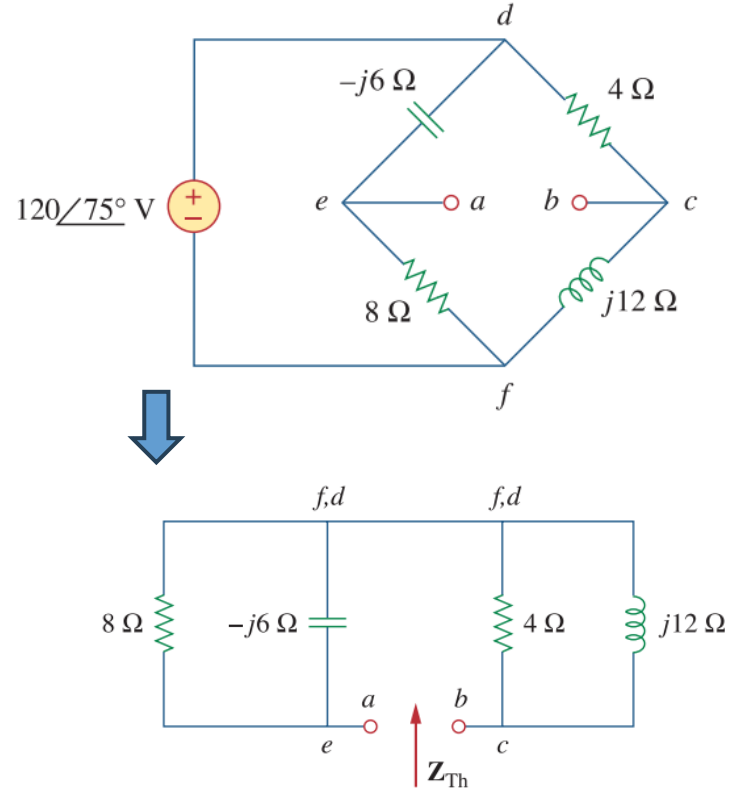
Solution

1- For  $Z_{th}$

$$\mathbf{Z}_1 = -j6 \parallel 8 = \frac{-j6 \times 8}{8 - j6} = 2.88 - j3.84 \Omega$$

$$\mathbf{Z}_2 = 4 \parallel j12 = \frac{j12 \times 4}{4 + j12} = 3.6 + j1.2 \Omega$$

$$\mathbf{Z}_{Th} = \mathbf{Z}_1 + \mathbf{Z}_2 = 6.48 - j2.64 \Omega$$



# Example 10.8- Thevenin Equivalent Circuit

Obtain the Thevenin equivalent at terminals a-b of the following circuit.

Solution (continue)

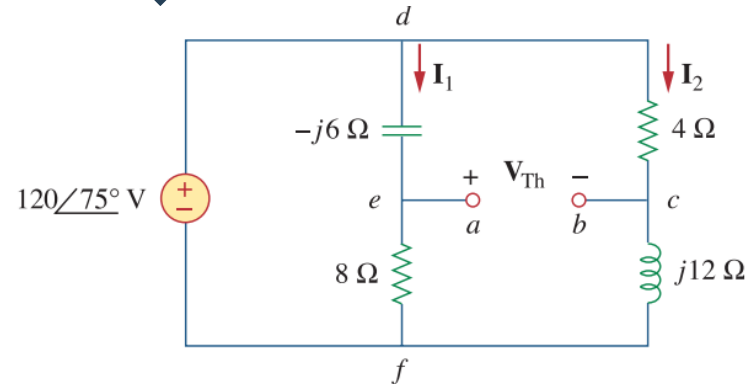
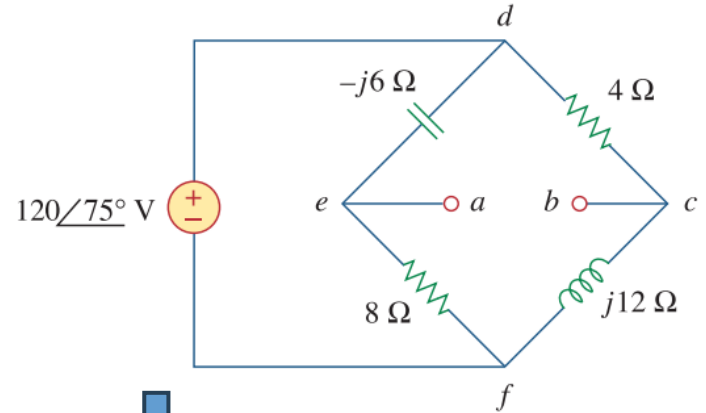
2- For  $V_{th}$

$$\mathbf{I}_1 = \frac{120 \angle 75^\circ}{8 - j6} \text{ A}, \quad \mathbf{I}_2 = \frac{120 \angle 75^\circ}{4 + j12} \text{ A}$$

Applying KVL around loop  $bcdeab$

$$\mathbf{V}_{Th} - 4\mathbf{I}_2 + (-j6)\mathbf{I}_1 = 0$$

$$\begin{aligned} \mathbf{V}_{Th} = 4\mathbf{I}_2 + j6\mathbf{I}_1 &= \frac{480 \angle 75^\circ}{4 + j12} + \frac{720 \angle 75^\circ + 90^\circ}{8 - j6} \\ &= 37.95 \angle 3.43^\circ + 72 \angle 201.87^\circ \\ &= -28.936 - j24.55 = 37.95 \angle 220.31^\circ \text{ V} \end{aligned}$$



# Example 10.9- Thevenin Equivalent Circuit

Obtain the Thevenin equivalent at terminals a-b of the following circuit.

## Solution

### 1- For $V_{th}$

To find  $V_{Th}$ , we apply KCL at node 1

$$15 = I_o + 0.5I_o \quad \Rightarrow \quad I_o = 10 \text{ A}$$

Applying KVL to the loop on the right-hand side

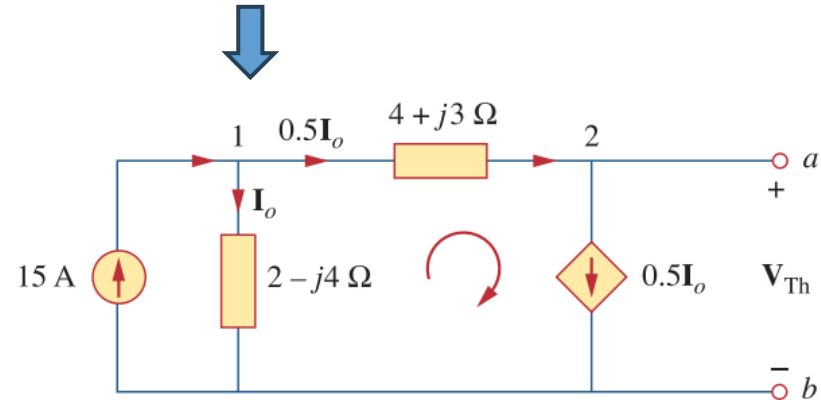
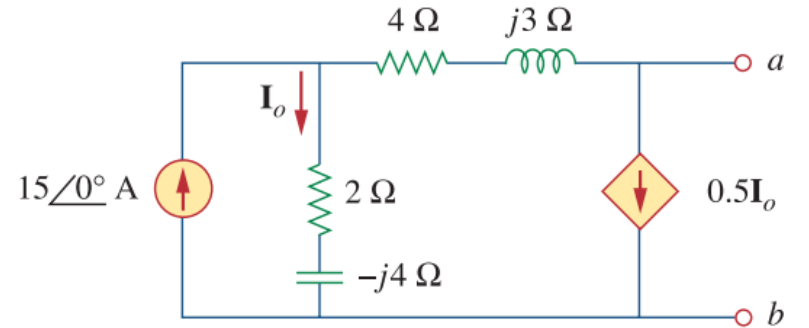
$$-I_o(2 - j4) + 0.5I_o(4 + j3) + V_{Th} = 0$$

or

$$V_{Th} = 10(2 - j4) - 5(4 + j3) = -j55$$

Thus, the Thevenin voltage is

$$V_{Th} = 55 \angle -90^\circ \text{ V}$$



# Example 10.9- Thevenin Equivalent Circuit

Obtain the Thevenin equivalent at terminals a-b of the following circuit.

Solution (continue)

2- For  $Z_{th}$

To obtain  $Z_{Th}$ , we remove the independent source. Due to the presence of the dependent current source, we connect a 3-A current source (3 is an arbitrary value chosen for convenience here, a number divisible by the sum of currents leaving the node) to terminals  $a-b$

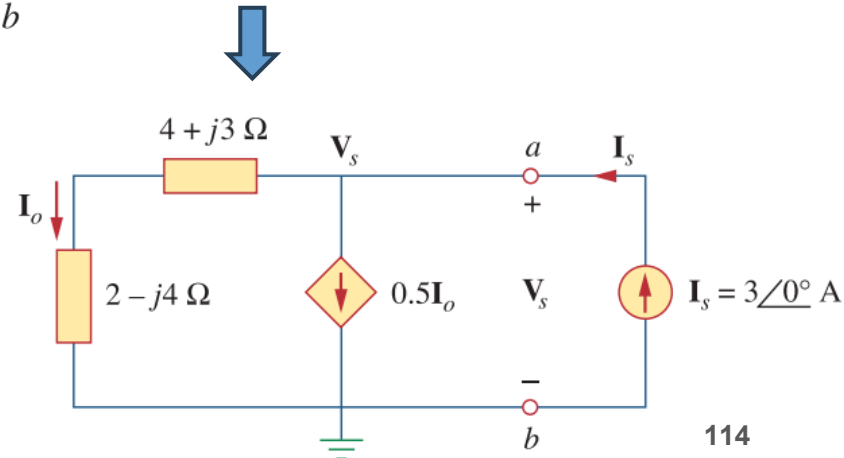
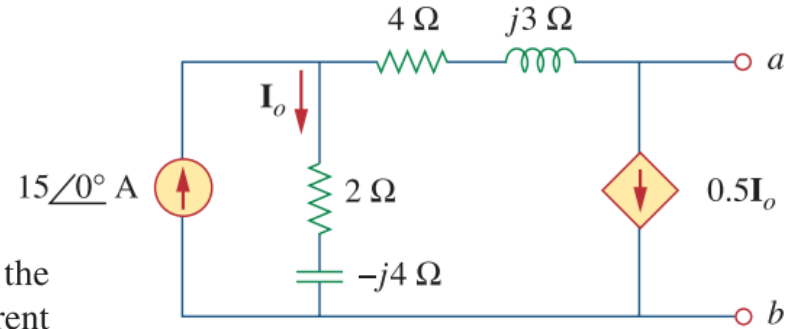
$$3 = \mathbf{I}_o + 0.5\mathbf{I}_o \quad \Rightarrow \quad \mathbf{I}_o = 2 \text{ A}$$

Applying KVL to the outer loop in Fig. 10.26(b) gives

$$\mathbf{V}_s = \mathbf{I}_o(4 + j3 + 2 - j4) = 2(6 - j)$$

The Thevenin impedance is

$$\mathbf{Z}_{Th} = \frac{\mathbf{V}_s}{\mathbf{I}_s} = \frac{2(6 - j)}{3} = 4 - j0.6667 \Omega$$



# Trigonometric Identities



- Sine and cosine form conversions.

$$\sin(A \pm B) = \sin A \cos B \pm \cos A \sin B$$

$$\cos(A \pm B) = \cos A \cos B \mp \sin A \sin B$$

$$\sin(\omega t \pm 180^\circ) = -\sin \omega t$$

$$\cos(\omega t \pm 180^\circ) = -\cos \omega t$$

$$\sin(\omega t \pm 90^\circ) = \pm \cos \omega t$$

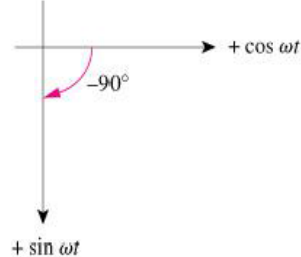
$$\cos(\omega t \pm 90^\circ) = \mp \sin \omega t$$

$$A \cos \omega t + B \sin \omega t = C \cos(\omega t - \theta)$$

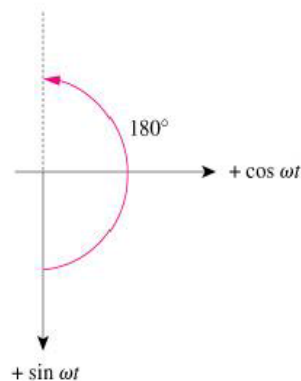
Where

$$C = \sqrt{A^2 + B^2} \quad \text{and} \quad \theta = \tan^{-1} \frac{B}{A}$$

Graphically relating sine and cosine functions.



$$\cos(\omega t - 90^\circ) = \sin \omega t$$



$$\sin(\omega t + 180^\circ) = -\sin \omega t$$



End of Lecture



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Questions?