

3- Circuit Theorems

Introduction and Linearity property

Superposition

Source transformations

Thevenin's theorem

Norton's theorem

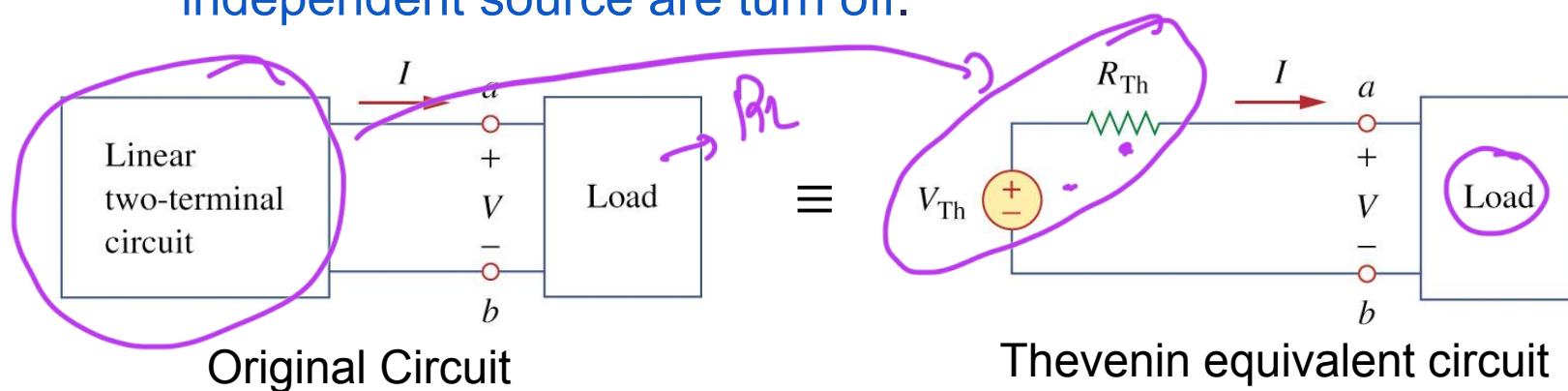
Maximum power transfer



Thevenin's Theorem

- Thevenin's theorem states that a linear two-terminal circuit can be replaced by an equivalent circuit consisting of a voltage source V_{Th} in series with a resistor R_{Th}

Where; V_{Th} is the open circuit voltage at the terminals and R_{Th} is the input or equivalent resistance at the terminals when the independent source are turn off.



Circuits

All ind

$$V_{th} = V_{OC}$$

$R_{th} \rightarrow$ Turn off all source

Volt \rightarrow short
current \rightarrow open

$$R_{eq} = R_{th}$$

$$= +V_{OC}$$



$$\overbrace{\quad\quad\quad}^{\text{Ind / dep}}$$

$$V_{th} = V_{OC}$$

$$R_{th} \rightarrow$$

Turn off all
ind source

$$R_{th} = \frac{V_{OC}}{I_{SC}}$$

$$R_{th} = \frac{V}{I_0}$$

$$= -V_{OC}$$



$$\overbrace{\quad\quad\quad}^{\text{All dep}}$$

$$V_{th} \rightarrow 0$$



$$R_{th} = \frac{V}{I}$$

التي تُولَّد، يُطلق عليه عادي (عادي)

Power التي تُولَّد في المُنْظَر "التي تُولَّد"

generated power /

diss power

$$P = I^2 R$$

$P + \rightarrow$ absorbed $= V^2 / R$

$P - \rightarrow$ emitted \rightarrow نَفَّع

Thevenin's equivalent circuit

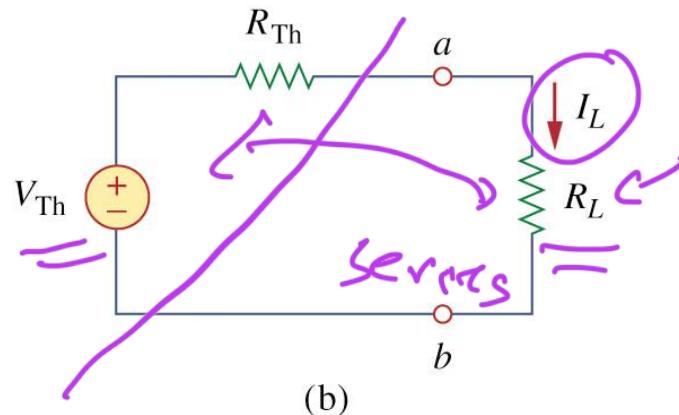
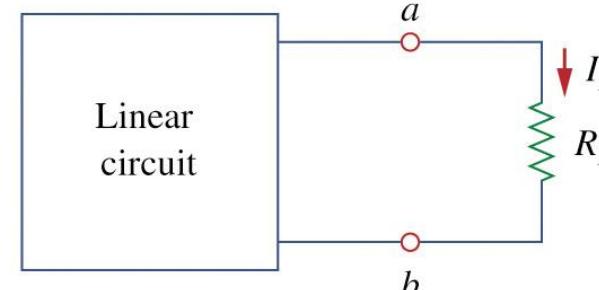
Simplified circuit

$$I_L = \frac{V_{Th}}{R_{Th} + R_L}$$

$$V_L = R_L I_L = \frac{R_L}{R_{Th} + R_L} V_{Th}$$

→ voltage divider

$$V_L = \frac{R_L}{R_L + R_{Th}} \times V_{Th}$$

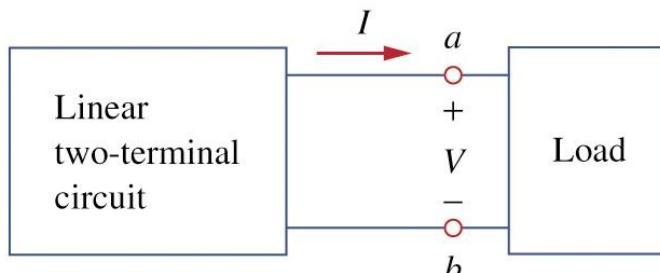


How to Find Thevenin Equivalent Circuit

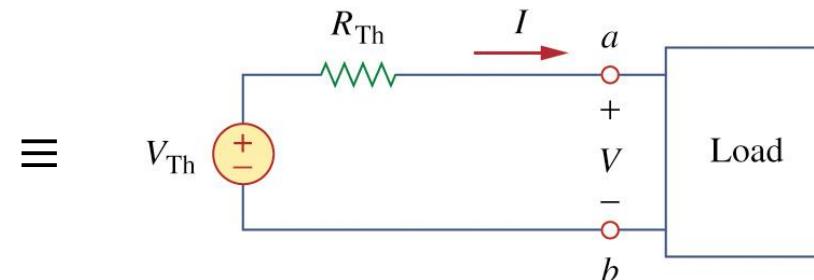
- First, **Open the circuit** (remove the load) at the **points** of interest **a-b**

1- V_{th} = Open circuit voltage (keep all sources intact)

2- R_{th} = Open circuit **equivalent resistance** appears at terminals **a-b** while all **independent sources=0** (voltage source=SC , current source=OC).



Original Circuit



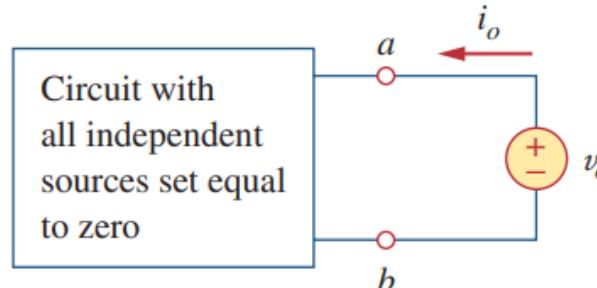
Thevenin equivalent circuit

How to Find Thevenin Equivalent Circuit

- ❖ To find R_{th} we have to consider two cases

- 1- Case 1 (Circuit has No dependent sources): turn off all independent sources, then find the equivalent R_{th} using series/parallel combinations.
- 2- Case 2 (Circuit has dependent sources): turn off all independent sources and keep dependent sources intact (as superposition). In order to find equivalent R_{th} , apply v_o and find the produced current i_o . Then, $R_{th} = \frac{v_o}{i_o}$

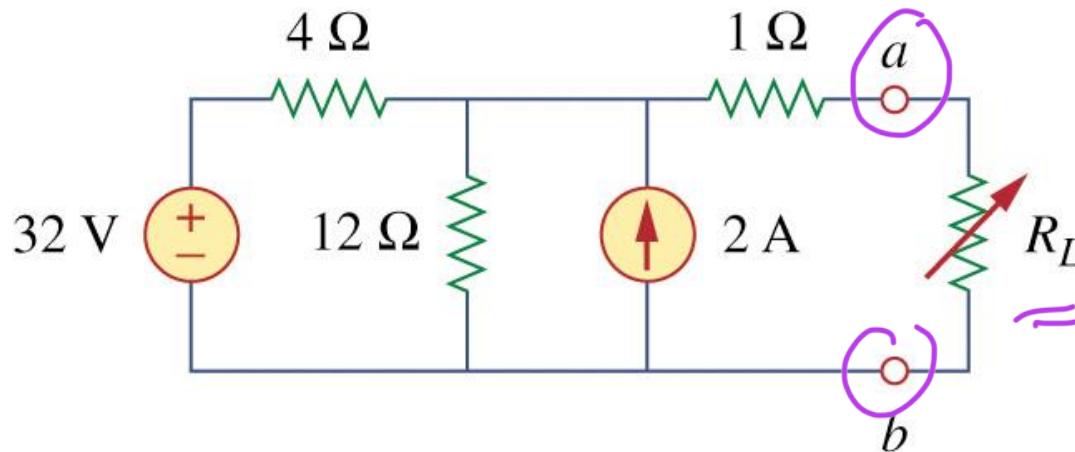
For simplicity: put $v_o = 1V$ and find the produced current i_o . Then, $R_{th} = \frac{1}{i_o}$



$$R_{Th} = \frac{v_o}{i_o}$$

Example 10

□ Find the Thevenin's equivalent of the circuit shown below, to the left of the terminals $a-b$. Then find the current through $R_L = 6, 16, \text{ and } 36 \Omega$.





node 1

analysis

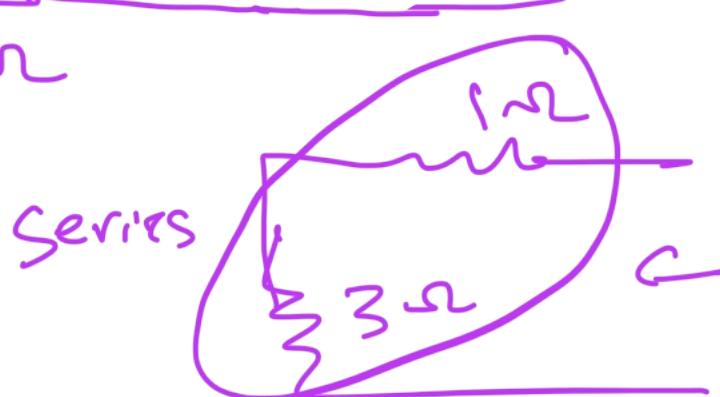
$$\begin{aligned}
 & \cancel{V = 32} \\
 & + \frac{N_f 0}{12} \quad \cancel{-2} = 0 \quad 2 + \frac{32}{u}
 \end{aligned}$$

$$\begin{aligned}
 & \cancel{= 2 \times \frac{1}{3} u} \quad V = 10 \times 3 \quad \rightarrow \boxed{V = 30V = V_{th}}
 \end{aligned}$$

R_{th} Thevenin \rightarrow Turn off



$$\frac{1}{\frac{1}{12} + \frac{1}{1}} = 3 \Omega$$



$$R_{th} = 1 + 3 = \boxed{4 \Omega}$$

Example 10

Solution:

For R_{th} ,

Voltage source = SC,
Current source = OC

$$R_{th} = 1 + 4//12 = 4\Omega$$

For V_{th} ,

Keep original circuit

$$V_{th} = V_{ac} \text{ (OC)}$$

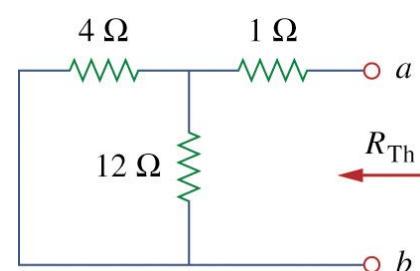
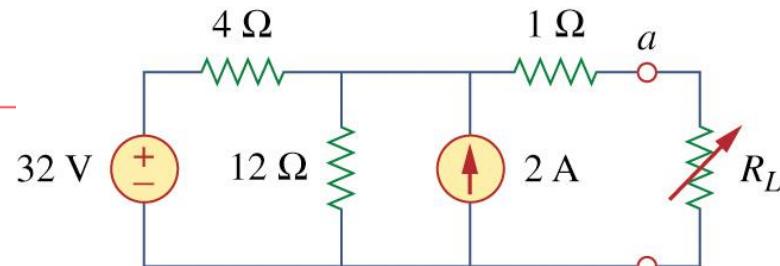
$$V_{th} = 12(i_1 - i_2) = ?$$

Using mesh analysis

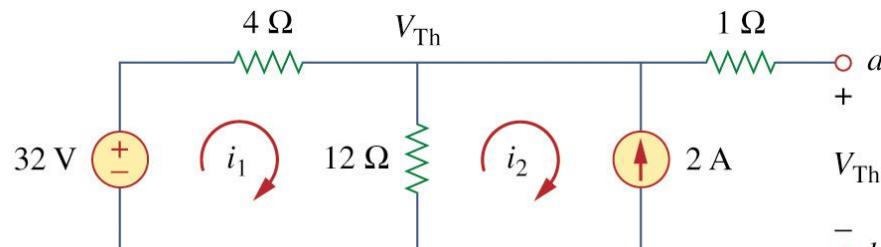
$$i_2 = -2A,$$

$$16i_1 - 12i_2 = 32,$$

$$i_1 = 0.5A,$$



(a)



(b)

$$V_{th} = 12(i_1 - i_2) = 30V$$

Example 10

Solution:

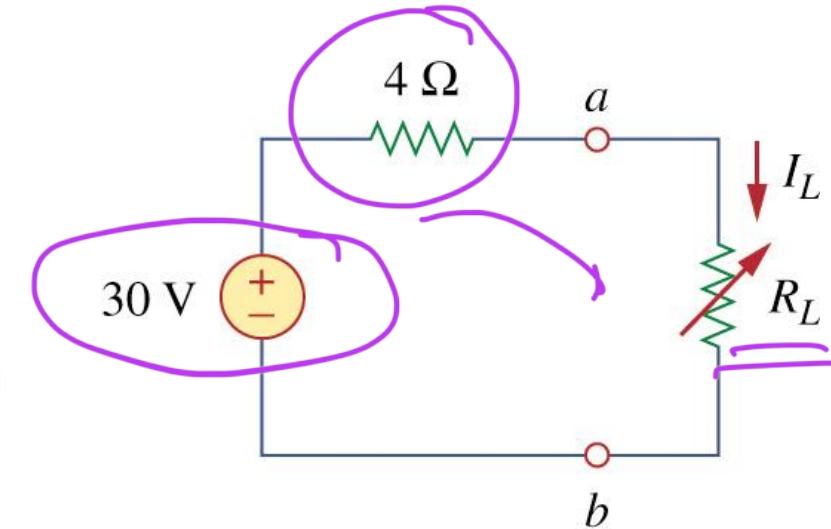
To get i_L :

$$i_L = \frac{V_{\text{Th}}}{R_{\text{Th}} + R_L} = \frac{30}{4 + R_L}$$

$$R_L = 6 \rightarrow i_L = 30/10 = 3\text{A}$$

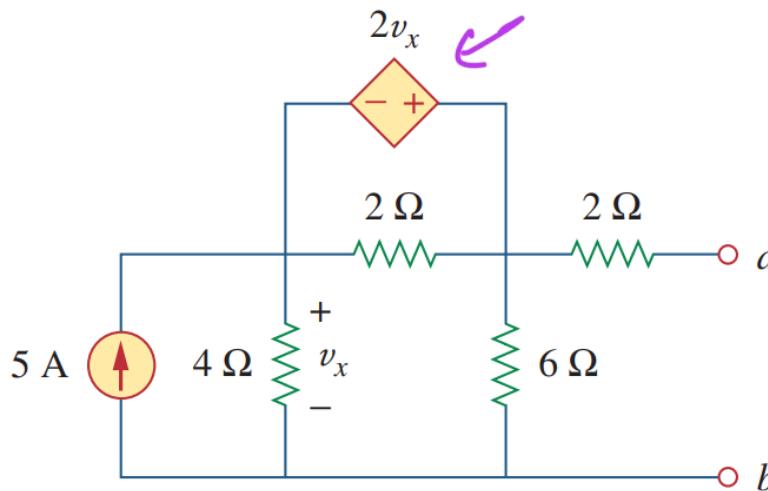
$$R_L = 16 \rightarrow i_L = 30/20 = 1.5\text{A}$$

$$R_L = 36 \rightarrow i_L = 30/40 = 0.75\text{A}$$



Example 11

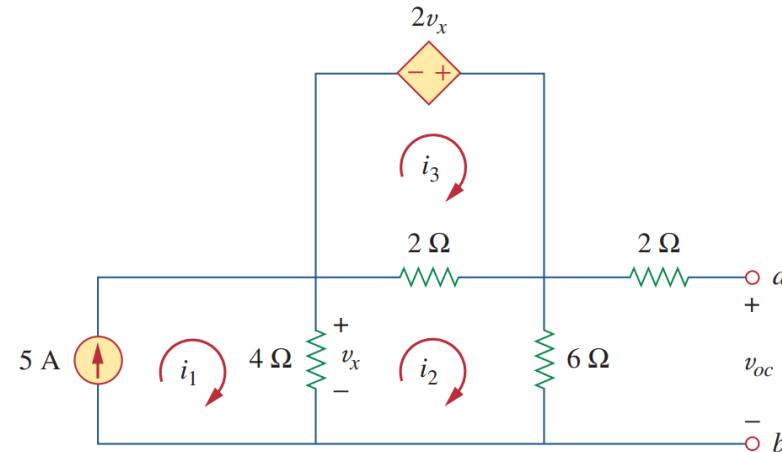
Find the Thevenin equivalent of the following circuit at terminals a-b.



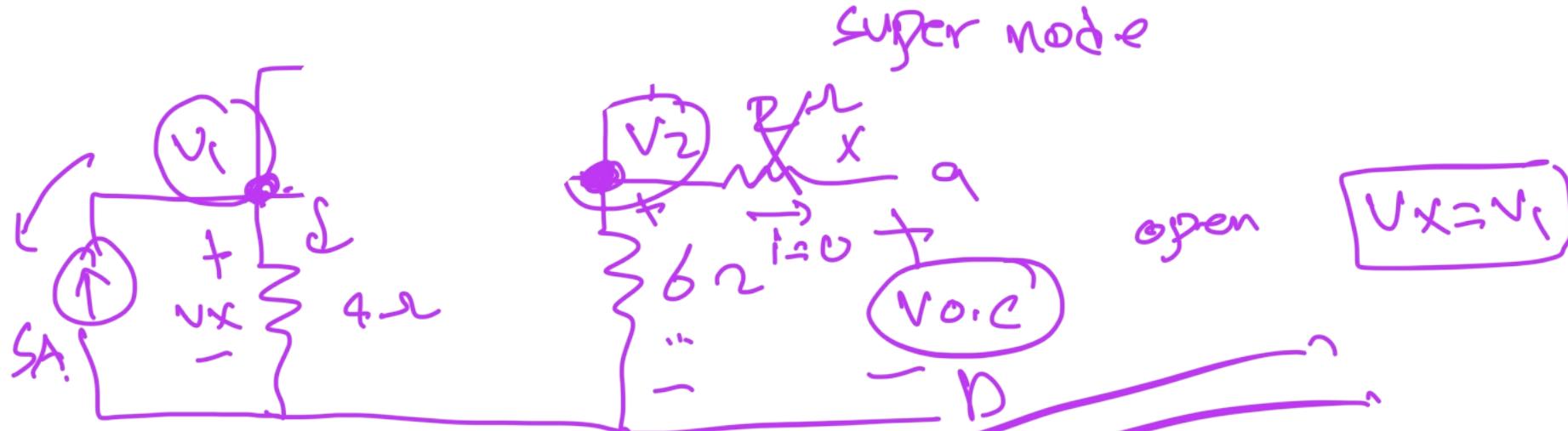
Solution:



1- For V_{th} , solve the following circuit to find v_{ab} (v_{oc}).



(b)



Nodal \leftarrow

$$v_2 - v_1 = 2v_x = 2v_1 + v_2$$

$$\Rightarrow 3v_1 - v_2 = 0 \quad \boxed{1}$$

node $1+2$

$$-5 + \frac{v_1 - 0}{4} + \frac{v_2 - 0}{6} = 0$$

$$\boxed{\frac{V_1}{6} + \frac{V_2}{6} = 5} \rightarrow ②$$

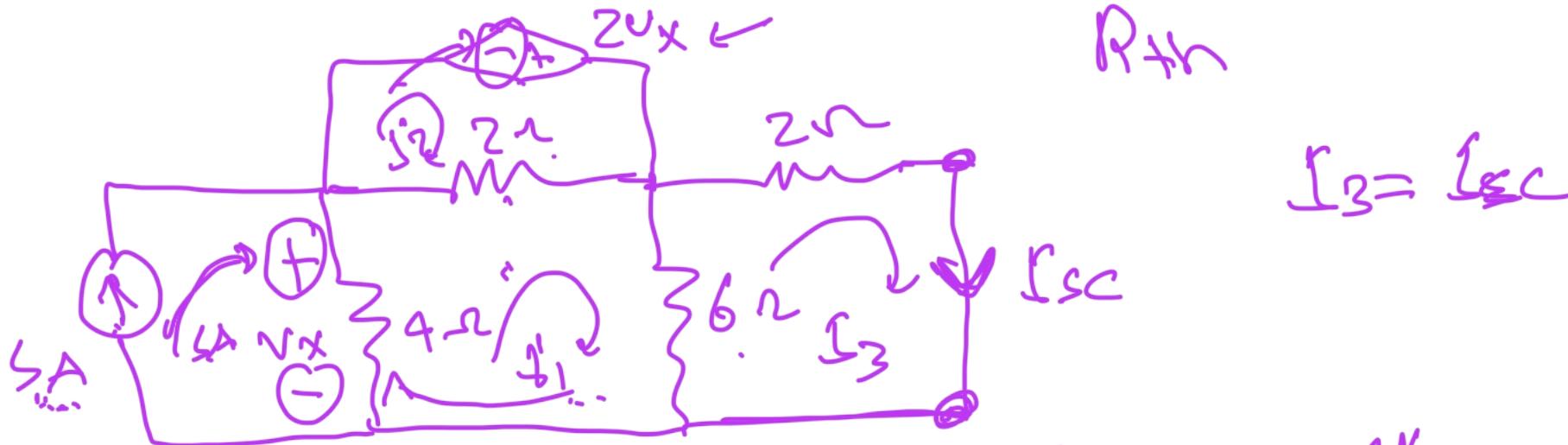
$$V_1 = \frac{20}{3} \text{ V}$$

mode $\rightarrow S \rightarrow 1$

$$V_2 = 20 \text{ V}$$

\Rightarrow

$$\boxed{V_{o.c} = V_{th} = V_2 = 20 \text{ V}}$$



$R + jX$

$$I_3 = I_{sc}$$

mesh

$$V_x = 4 \overrightarrow{(S - I_1)} = 20 - 4I_1$$

mesh ①

$$12I_1 - 2I_2 - 6I_3 - 5 \times 4 = 0$$

$$\Rightarrow \boxed{12I_1 - 2I_2 - 6I_3 = 20} \rightarrow ①$$

mesh ②

$$-2f_1 + 2f_2 = 2(20 - 4f_1) = 0$$

$$-2f_1 + 2f_2 = 40 + 8f_1 = 0$$

$$6f_1 + 2f_2 + 0f_3 = 40 \rightarrow ②$$

mesh ③

$$-6f_1 + 0f_2 + 8f_3 = 0 \rightarrow ③$$

mode $\rightarrow S \rightarrow 2$

$$I_1 = \frac{40}{2} \quad , \quad I_2 = \frac{20}{3} \quad , \quad I_3 = \frac{10}{3}$$

$$I_{SC} = I_3 = \frac{10}{3} A$$

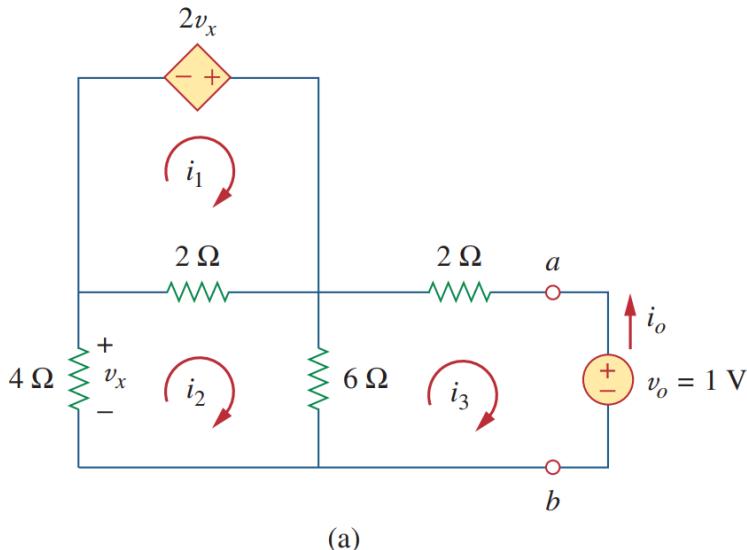
$$R_{th} = \frac{V_o.c}{I_{SC}} = \frac{20}{10/3} = 6 \Omega$$

Example 11

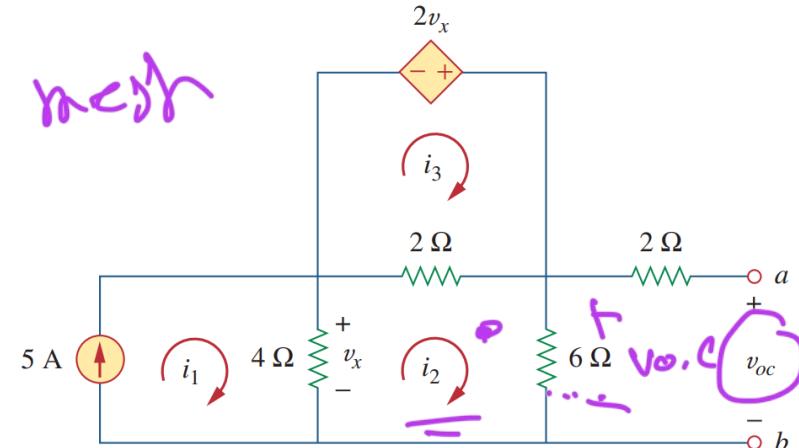
Find the Thevenin equivalent of the following circuit at terminals a-b.

Solution:

2- For R_{th} , solve the following circuit to find i_o . Then, $R_{th} = 1/i_o$



1- For V_{th} , solve the following circuit to find v_{ab} (v_{oc}).



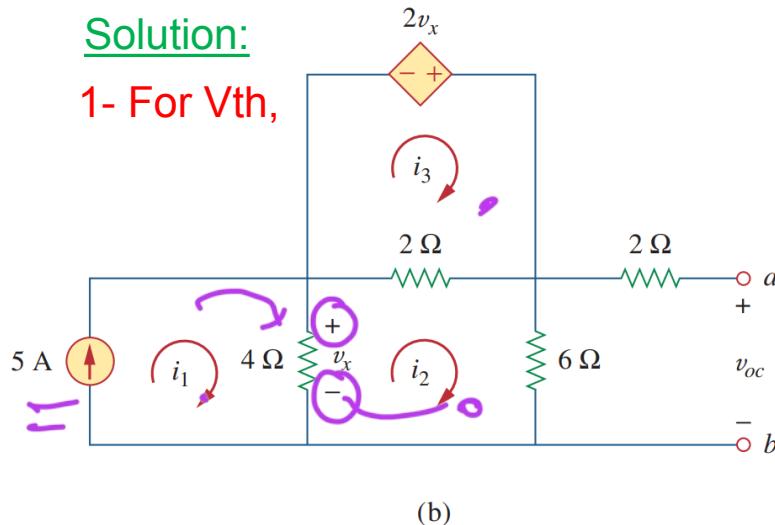
$$V_{oc} = \sum R_{eq}$$

Example 11

Find the Thevenin equivalent of the following circuit at terminals a-b.

Solution:

1- For V_{th} ,



Applying mesh analysis:

$$\text{Mesh 1: } i_1 = 5$$

$$\text{Mesh 2: } 4(i_2 - i_1) + 2(i_2 - i_3) + 6i_2 = 0$$

$$\Rightarrow 12i_2 - 4i_1 - 2i_3 = 0$$

$$\text{Mesh 3: } -2v_x + 2(i_3 - i_2) = 0 \Rightarrow v_x = i_3 - i_2$$

$$\text{But, } 4(i_1 - i_2) = v_x \Rightarrow 4i_1 - 3i_2 - i_3 = 0$$

Solving these equations leads to $i_2 = 10/3$. ✓

$$V_{th} = v_{oc} = 6i_2 = 20 \text{ V}$$

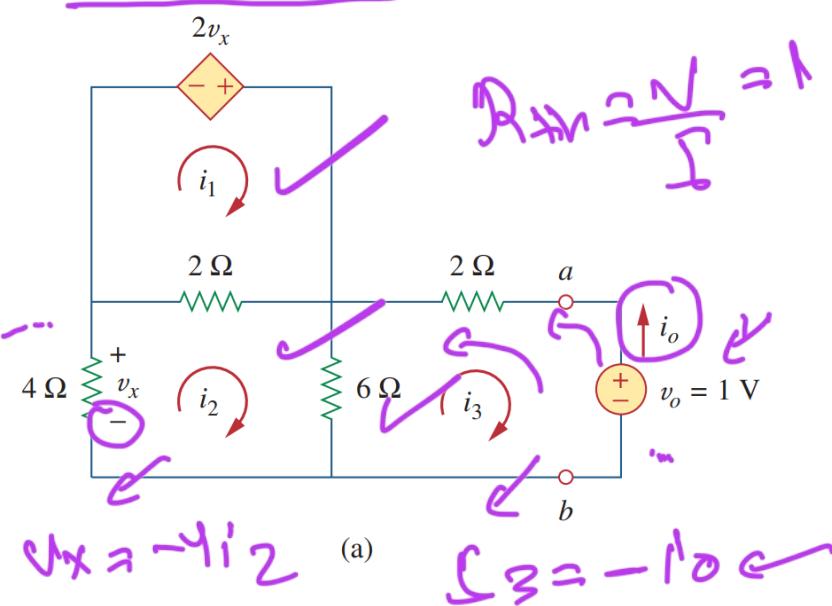
Example 11

Find the Thevenin equivalent of the following circuit at terminals a-b.

Solution:

2- For R_{th} , Put $v_o = 1V$

and kill any independent sources



Applying mesh analysis:

$$\text{Mesh 1: } -2v_x + 2(i_1 - i_2) = 0 \quad \text{or} \quad v_x = i_1 - i_2$$

$$\text{But } -4i_2 = v_x = i_1 - i_2; \text{ hence,} \Rightarrow i_1 = -3i_2$$

$$\text{Mesh 2: } 4i_2 + 2(i_2 - i_1) + 6(i_2 - i_3) = 0$$

$$\text{Mesh 3: } 6(i_3 - i_2) + 2i_3 + 1 = 0$$

Solving these equations gives

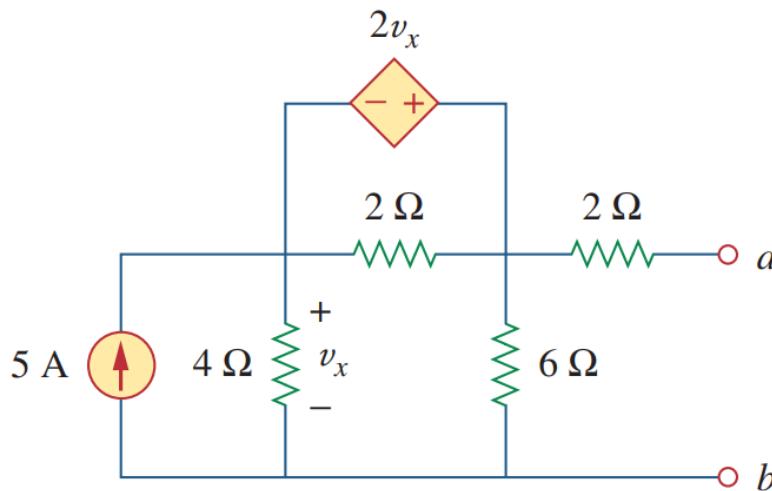
$$i_3 = -\frac{1}{6} \text{ A}$$

$$\text{But } i_o = -i_3 = \frac{1}{6} \text{ A. Hence,}$$

$$R_{th} = \frac{1 \text{ V}}{i_o} = 6 \Omega$$

Example 11

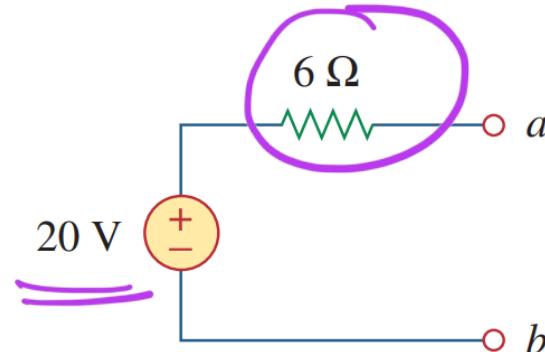
Find the Thevenin equivalent of the following circuit at terminals a-b.



Solution:



≡



3- Circuit Theorems 2

Introduction and Linearity property

Superposition

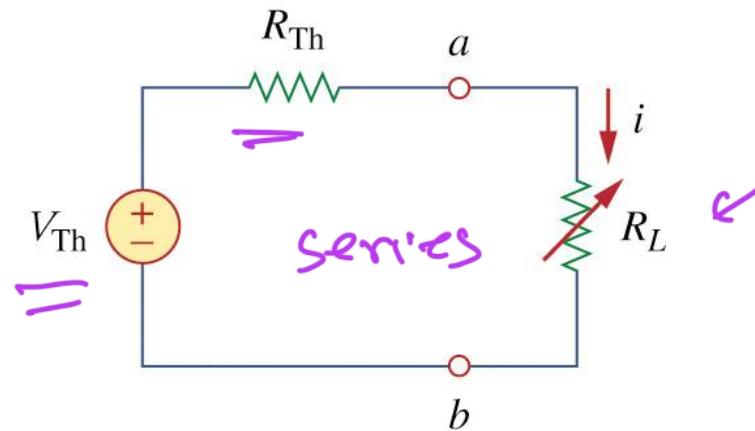
Source transformations

Thevenin's theorem

Norton's theorem

Maximum power transfer

Maximum Power Transfer



$\rightarrow R_L \rightarrow \text{max power}$

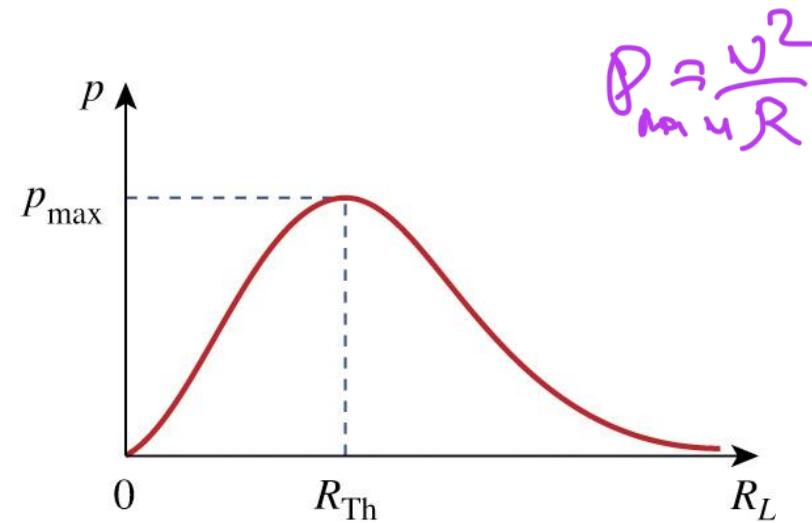
Find max power

$$p = i^2 R_L = \left(\frac{V_{TH}}{R_{TH} + R_L} \right)^2 R_L$$

- Maximum power is transferred to the load when the load resistance equals the Thevenin resistance as seen the load ($R_L = R_{TH}$).

$$R_L = R_{TH}$$

$$p_{\max} = \frac{V_{TH}^2}{4R_{TH}}$$



Maximum Power Transfer: Mathematical Proof

$$\frac{dp}{dR_L} = V_{TH}^2 \left[\frac{(R_{TH} + R_L)^2 - 2R_L(R_{TH} + R_L)}{(R_{TH} + R_L)^4} \right]$$

$$= V_{TH}^2 \left[\frac{(R_{TH} + R_L - 2R_L)}{(R_{TH} + R_L)^3} \right] = 0$$

$$0 = (R_{TH} + R_L - 2R_L) = (R_{TH} - R_L)$$

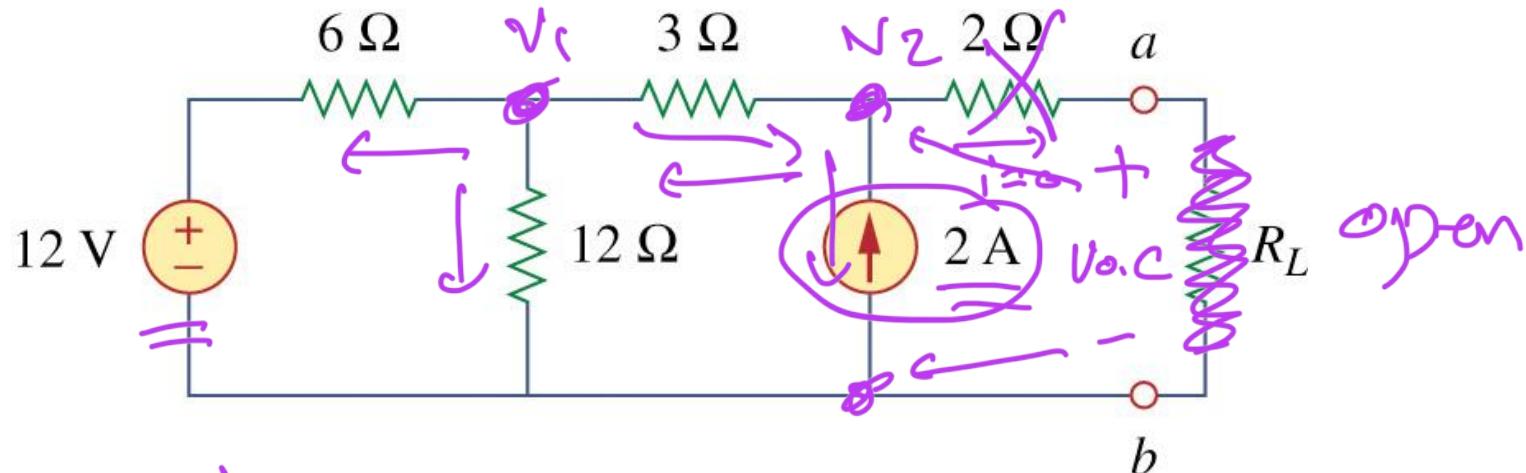
$$R_L = R_{TH}$$

$$p_{\max} = \frac{V_{TH}^2}{4R_{TH}}$$

calc ①

Example 12

□ Find the value of R_L for maximum power transfer in the circuit of the following figure. Find the maximum power.



Nodal Analysis

$$\frac{V_2 - V_1}{12} + \frac{V_1}{12} + \frac{V_1 - V_2}{3} = 0$$

$$\frac{V_2 - V_1}{3} + -2 = 0$$

$$\rightarrow \boxed{\frac{V_2 - V_1}{12} + \frac{V_1}{12} = 2} \rightarrow ①$$

$$\boxed{\frac{-V_1}{3} + \frac{V_2}{3} = 2} \rightarrow ②$$

$$V_1 = 16 \text{ V}$$

$$V_2 = 22 \text{ V}$$

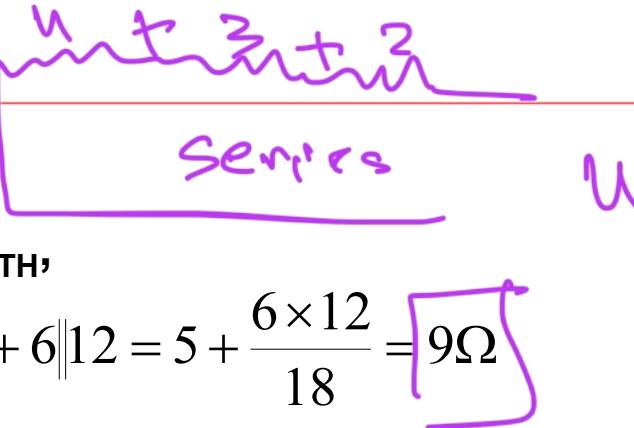
$= V_{th}$

Example 12

Solution:

1- To find R_{TH} ,

$$R_{TH} = 2 + 3 + 6\parallel 12 = 5 + \frac{6 \times 12}{18} = 9\Omega$$

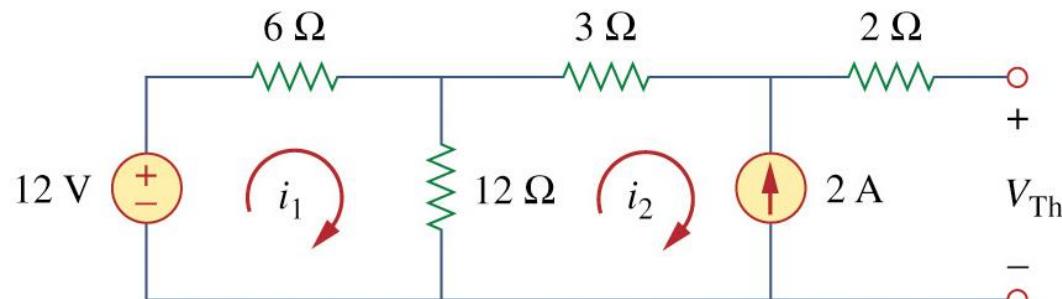


2- To find V_{TH} ,

$$\text{Mesh: } -12 + 18i_1 - 12i_2, \quad i_2 = -2A$$

$$-12 + 6i_1 + 3i_2 + 2(0) + V_{TH} = 0$$

$$\therefore V_{TH} = 22V$$



For maximum power:

$$R_L = R_{TH} = 9\Omega$$

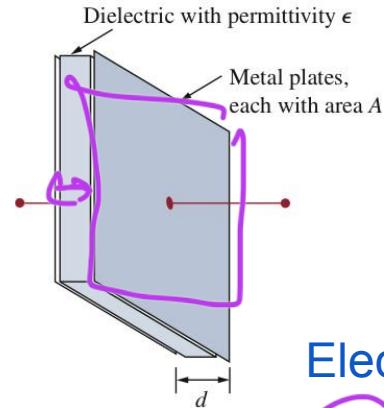
$$P_{max} = \frac{V_{TH}^2}{4R_L} = \frac{22^2}{4 \times 9} = 13.44W$$



4- Capacitors and Inductors



Summary of Capacitors



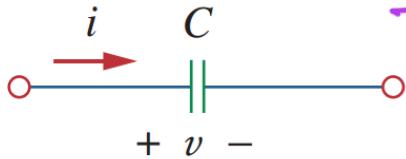
Physical structure

$$C = \frac{\epsilon A}{d}$$

Electrical Characteristics

$$q = Cv$$

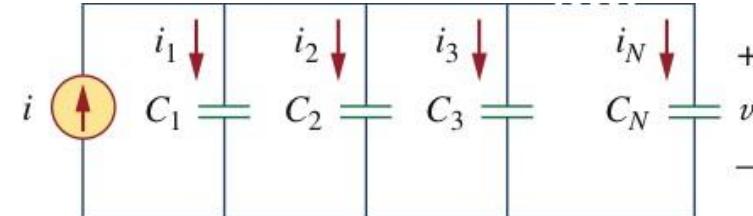
$$i = C \frac{dv}{dt}$$



Energy Stored

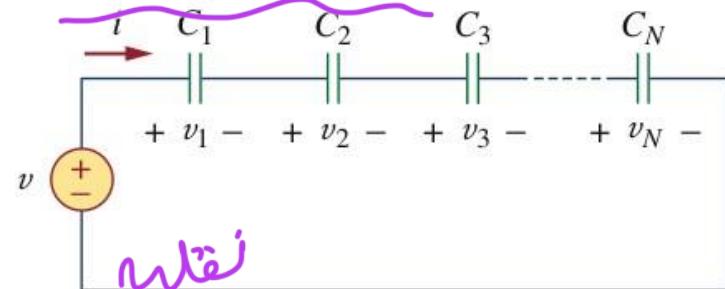
$$w = \frac{1}{2} C v^2$$

Parallel Capacitors



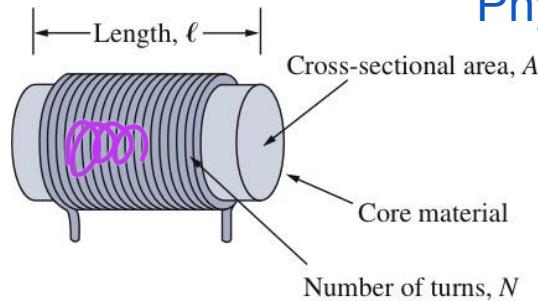
$$C_{eq} = C_1 + C_2 + C_3 + \dots + C_N$$

Series Capacitors



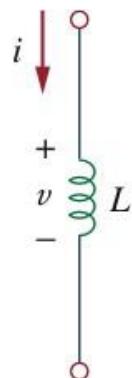
$$\frac{1}{C_{eq}} = \frac{1}{C_1} + \frac{1}{C_2} + \frac{1}{C_3} + \dots + \frac{1}{C_N}$$

Summary of Inductors



Physical structure

$$L = \frac{N^2 \mu A}{l}$$



Electrical Characteristics

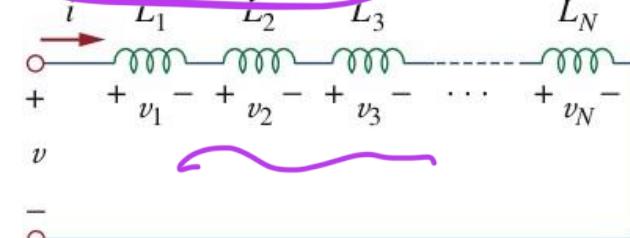
$$v = L \frac{di}{dt}$$

S.C. in DC

Energy Stored

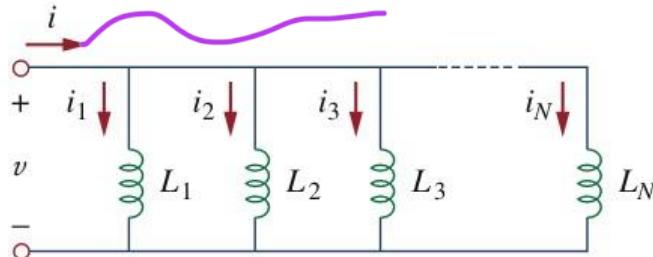
$$w(t) = \frac{1}{2} L i^2(t)$$

Series Inductors



$$L_{eq} = L_1 + L_2 + L_3 + \dots + L_N$$

Parallel Inductors



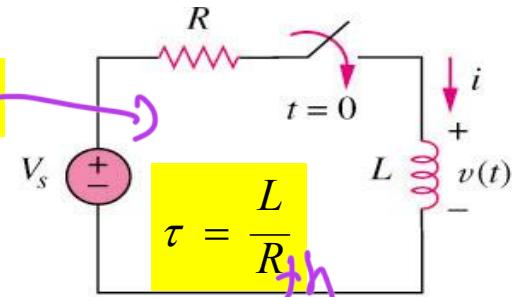
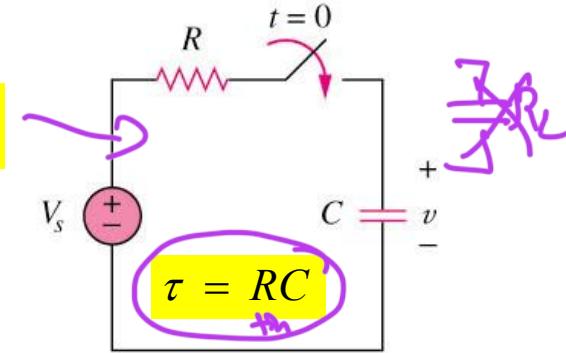
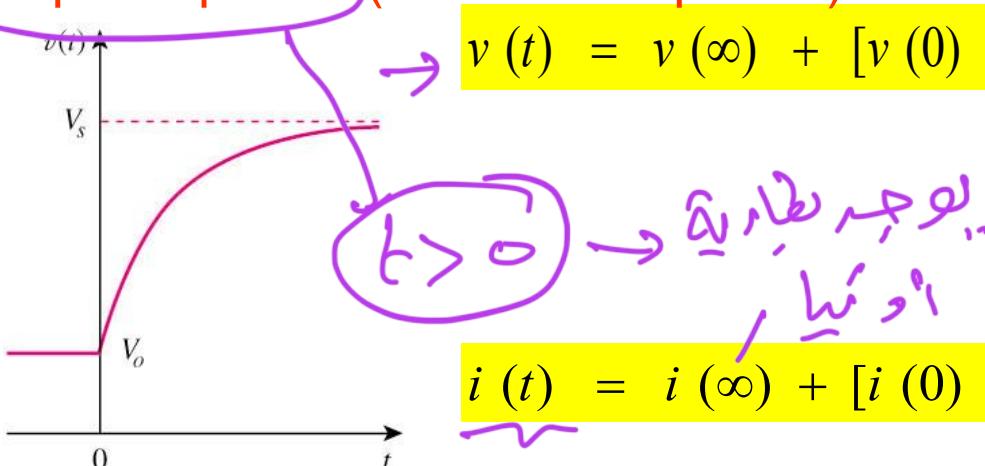
$$\frac{1}{L_{eq}} = \frac{1}{L_1} + \frac{1}{L_2} + \dots + \frac{1}{L_N}$$

Remember: Important characteristics of basic elements

Relation	Resistor (R)	Capacitor (C)	Inductor (L)
v - i :	$v = iR$	$v = \frac{1}{C} \int_{t_0}^t i(\tau) d\tau + v(t_0)$	$v = L \frac{di}{dt}$
i - v :	$i = v/R$	$i = C \frac{dv}{dt}$	$i = \frac{1}{L} \int_{t_0}^t v(\tau) d\tau + i(t_0)$
p or w :	$p = i^2R = \frac{v^2}{R}$	$w = \frac{1}{2} Cv^2$	$w = \frac{1}{2} Li^2$
Series:	$R_{\text{eq}} = R_1 + R_2$	$C_{\text{eq}} = \frac{C_1 C_2}{C_1 + C_2}$	$L_{\text{eq}} = L_1 + L_2$
Parallel:	$R_{\text{eq}} = \frac{R_1 R_2}{R_1 + R_2}$	$C_{\text{eq}} = C_1 + C_2$	$L_{\text{eq}} = \frac{L_1 L_2}{L_1 + L_2}$
At dc:	Same	Open circuit	Short circuit
Circuit variable that cannot change abruptly:	Not applicable	v	i

Summary of Transient-Response for RC and RL circuits

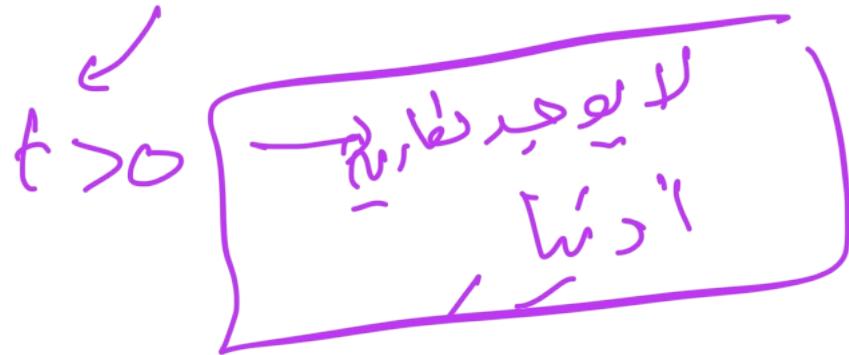
Step Response (Forced Response):



Three steps to solve any transient-response problem:

1. The initial Values. $v(0), i(0) \leftarrow t < 0$
2. The final value (after long time, C acts as OC and L acts as SC). $v(\infty), i(\infty)$
3. The time constant $\tau \rightarrow t > 0$

Note: Natural Response (source free) is a special case where final value = 0



End of Lecture

Questions?